#### Electromechanical System Dynamics, energy Conversion, and Electromechanical Analogies

#### Modeling of Dynamic Systems

Modeling of dynamic systems may be done in several ways:

- Use the standard equation of motion (Newton's Law) for mechanical systems
- Use circuits theorems (Ohm's law and Kirchhoff's laws: KCL and KVL)
- Another approach utilizes the notation of energy to model the dynamic system (Lagrange model).

#### Mathematical Modeling and System Dynamics Newtonian Mechanics: Translational Motion

- The equations of motion of mechanical systems can be found using Newton's second law of motion. F is the vector sum of all forces applied to the body; a is the vector of acceleration of the body with respect to an inertial reference frame; and m is the mass of the body.
- To apply Newton's law, the free-body diagram in the coordinate system used should be studied.

$$\sum \mathbf{F} = m\mathbf{a}$$

#### Translational Motion in Electromechanical Systems

- Consideration of friction is essential for understanding the operation of electromechanical systems.
- Friction is a very complex nonlinear phenomenon and is very difficult to model friction.
- The classical Coulomb friction is a retarding frictional force (for translational motion) or torque (for rotational motion) that changes its sign with the reversal of the direction of motion, and the amplitude of the frictional force or torque are constant.
- Viscous friction is a retarding force or torque that is a linear function of linear or angular velocity.

#### Newtonian Mechanics: Translational Motion

For one-dimensional rotational systems, Newton's second law of motion is expressed as the following equation. *M* is the sum of all moments about the center of mass of a body (N-m); *J* is the moment of inertial about its center of mass (kg/m<sup>2</sup>); and α is the angular acceleration of the body (rad/s<sup>2</sup>).

 $M = j\alpha$ 

# The Lagrange Equations of Motion

- Although Newton's laws of motion form the fundamental foundation for the study of mechanical systems, they can be straightforwardly used to derive the dynamics of electromechanical motion devices because electromagnetic and circuitry transients behavior must be considered. This means, the circuit dynamics must be incorporated to find augmented models.
- This can be performed by integrating torsional-mechanical dynamics and sensor/actuator circuitry equations, which can be derived using Kirchhoff's laws.
- Lagrange concept allows one to integrate the dynamics of mechanical and electrical components. It employs the scalar concept rather the vector concept used in Newton's law of motion to analyze much wider range of systems than F =ma.
- With Lagrange dynamics, focus is on the entire system rather than individual components.
- $\Gamma$ , D,  $\Pi$  are the total kinetic, dissipation, and potential energies of the system.  $q_i$  and  $Q_i$  are the generalized coordinates and the generalized applied forces (input).

$$\frac{d}{dt} \left( \frac{d\Gamma}{\frac{1}{d q_i}} \right) - \frac{d\Gamma}{d q_i} + \frac{dD}{\frac{1}{d q_i}} + \frac{d\Pi}{\frac{1}{d q_i}} = Q_i$$

### **Electrical and Mechanical Counterparts**

Energy	Mechanical	Electrical
Kinetic	<b>Mass / Inertia</b> 0.5 <i>mv</i> <sup>2</sup> / 0.5 <i>jω</i> <sup>2</sup>	Inductor 0.5 <i>Li</i> <sup>2</sup>
Potential	<b>Gravity</b> : <i>mgh</i> <b>Spring:</b> 0.5 <i>kx</i> <sup>2</sup>	Capacitor 0.5 Cv <sup>2</sup>
Dissipative	<b>Damper / Friction</b> 0.5 <i>Bv</i> <sup>2</sup>	Resistor <i>Ri</i> <sup>2</sup>

#### Mathematical Model for a Simple Pendulum

The kinetic energy of the pendulum bob is :  $\Gamma = \frac{1}{2}mv^2 = \frac{1}{2}m\left(l\dot{\theta}\right)^2$ 

The potential energy is :  $\Pi = mgh = mgl(1 - \cos\theta)$ 



## **Electrical Conversion**



Energy Transfer in Electromechanical Systems

For rotational motion, the electromagnetic torque, as a function

of current and angular displacement, is :  $T_e(i,\theta) = \frac{dW_c(i,\theta)}{d\theta}$ 

Where 
$$W_c = \oint_i \psi di$$
; where  $\psi$  is the flux.

## **Electromechanical Analogies**

 From Newton's law or using Lagrange equations of motions, the secondorder differential equations of translational-dynamics and torsionaldynamics are found as

$$m\frac{d^{2}x}{dt^{2}} + B_{v}\frac{dx}{dt} + k_{s}x = F_{a}(t) \text{ (Translational dynamics)}$$
$$j\frac{d^{2}\theta}{dt^{2}} + B_{m}\frac{d\theta}{dt} + k_{s}\theta = T_{a}(t) \text{ (Torsional dynamics)}$$

For a series RLC circuit, find the characteristic equation and define the analytical relationships between the characteristic roots and circuitry parameters.

$$\frac{d^2i}{dt^2} + \frac{R}{L}\frac{di}{dt} + \frac{1}{LC}i = \frac{1}{L}\frac{dv_a}{dt}$$
$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

The characteristic roots are

$$s_{1} = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^{2} - \frac{1}{LC}}$$
$$s_{2} = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^{2} - \frac{1}{LC}}$$

## Resistance, R (ohm)

Appied voltage v(t)Current i(t)v(t) = Ri(t) $i(t) = \frac{1}{R}v(t)$ 



## Inductance, L (H)





## Capacitance, C (F)



$$v(t) = \frac{1}{C} \int_{t_0}^t i(t) dt$$
$$i(t) = C \frac{dv(t)}{dt}$$



# Translational Damper, $B_v$ (N-sec)

Appied force  $F_{a}(t)$  in Newton Linear velocity v(t) (m/sec) Linear position x(t) (m)  $F_a(t) = B_m v(t)$  $v(t) = \frac{1}{B_m} F_a(t)$  $F_a(t) = B_m v(t) = B_m \frac{dx(t)}{dt}$  $x(t) = \frac{1}{B_{y}} \int_{t}^{t} F_{a}(t) dt$ 



# Translational Spring, k (N)

Appied force  $F_a(t)$  in Newton Linear velocity v(t) (m/sec) Linear position x(t) (m)  $F_a(t) = k_s x(t)$  $x(t) = \frac{1}{k_s} F_a(t)$  $v(t) = \frac{dx(t)}{dt} = \frac{1}{k_s} \frac{dF_a(t)}{dt}$  $F_a(t) = k_s \int_{-\infty}^{t} v(t) dt$ 



## Rotational Damper, $B_m$ (N-m-sec/rad)

Appied torque  $T_{a}(t)$  (N - m) Angular velocity  $\omega(t)$  (rad/sec) Angular displacement  $\theta(t)$  (rad)  $T_a(t) = B_m \omega(t)$  $\omega(t) = \frac{1}{B_m} T_a(t)$  $T_a(t) = B_m \omega(t) = B_m \frac{d\theta(t)}{dt}$  $\theta(t) = \frac{1}{B_{m}} \int_{0}^{t} T_{a}(t) dt$ 



## Rotational Spring, k<sub>s</sub> (N-m-sec/rad)

Appied torque  $T_{a}(t)$  (N - m) Angular velocity  $\omega(t)$  (rad/sec) Angular displacement  $\theta(t)$  (rad)  $T_a(t) = B_m \theta(t)$  $\theta(t) = \frac{1}{k_s} T_a(t)$  $\omega(t) = \frac{d\theta(t)}{dt} = \frac{1}{k_s} \frac{dT_a(t)}{dt}$  $T_a(t) = k_s \int_{0}^{t} \omega(t) dt$ 



### Mass Grounded, m (kg)

Appied torque  $T_a(t)$  (N - m) Linear velocity v(t) (m/sec) Linear position x(t) (m)

$$F_a(t) = m \frac{dv}{dt} = m \frac{d^2 x(t)}{dt^2}$$
$$v(t) = \frac{1}{m} \int_{t_0}^t F_a(t) dt$$



## Mass Grounded, m (kg)

Appied torque  $T_{a}(t)$  (N - m) Angular velocity  $\omega(t)$  (rad/sec) Angular displacement  $\theta(t)$  (rad)

$$T_{a}(t) = J \frac{d\omega}{dt} = J \frac{d^{2}\theta(t)}{dt^{2}}$$
$$\omega(t) = \frac{1}{J} \int_{t0}^{t} T_{a}(t) dt$$



## Steady-State Analysis

- State: The state of a dynamic system is the smallest set of variables (called state variables) so that the knowledge of these variables at  $t = t_0$ , together with the knowledge of the input for  $t \ge t_0$ , determines the behavior of the system for any time  $t \ge t_0$ .
- **State Variables:** The state variables of a dynamic system are the variables making up the smallest set of variables that determine the state of the dynamic system.
- **State Vector:** If *n* state variables are needed to describe the behavior of a given system, then the *n* state variables can be considered the *n* components of a vector *x*. Such vector is called a state vector.
- **State Space:** The *n*-dimensional space whose coordinates axes consist of the  $x_1$  axis,  $x_2$  axis, ...,  $x_n$  axis, where  $x_1$ ,  $x_2$ , ...,  $x_n$  are state variables, is called a state space.
- State-Space Equations: In state-space analysis we are concerned with three types of variables that are involved in the modeling of dynamic system: input variables, output variables, and state variables.



The state variables describe the future response of a system, given the present state, the excitation inputs, and the equations describing the dynamics

## Electrical Example: An RLC Circuit

 $x_{1} = v_{C}(t); x_{2} = i_{L}(t)$   $\xi = (1/2)Li_{L}^{2} + (1/2)Cv_{c}^{2}$   $x_{1}(t_{0}) \text{ and } x_{2}(t_{0}) \text{ is the total initial}$ energy of the network USE KCL at the junction

$$i_c = C \frac{dv_c}{dt} = +u(t) - i_L$$



### The State Differential Equation

$$x_1 = a_{11}x_1 + a_{12}x_2 + \dots + a_{1n}x_n + b_{11}u_1 + \dots + b_{1m}u_m$$

$$x_2 = a_{21}x_1 + a_{22}x_2 + \dots + a_{2n}x_n + b_{21}u_1 + \dots + b_{2m}u_m$$

$$x_n = a_{n1}x_1 + a_{n2}x_2 + \dots + a_{nn}x_n + b_{n1}u_1 + \dots + b_{nm}u_m$$

State Vector

$$\frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \\ . \\ . \\ x_n \end{bmatrix} = \begin{bmatrix} a_{11} & a_{12} & a_{1n} \\ a_{21} & a_{22} & a_{2n} \\ . & . & . \\ a_{n1} & a_{n2} & a_{nn} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ . \\ . \\ x_n \end{bmatrix} + \begin{bmatrix} b_{11} \dots b_{1m} \\ . \dots \\ b_{n1} \dots b_{nm} \end{bmatrix} \begin{bmatrix} u_1 \\ . \\ u_m \end{bmatrix}$$

x = Ax + Bu (State Differential Equation)A : State matrix; B : input matrixy = Cx + Du (Output Equation)C : Output matrix; D : direct transmission matrix

### The Output Equation

$$y_{1} = h_{11}x_{1} + h_{12}x_{2} + \dots + h_{1n}x_{n}$$
  

$$y_{2} = h_{21}x_{1} + h_{22}x_{2} + \dots + h_{2n}x_{n}$$
  

$$y_{b} = a_{b1}x_{1} + a_{b2}x_{2} + \dots + a_{bn}x_{n}$$

$$y = \begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_b \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} & h_{1n} \\ h_{21} & h_{22} & h_{2n} \\ \vdots & \vdots & \vdots \\ h_{b1} & h_{b2} & h_{bn} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} = Hx$$

x = Ax + Bu (State Differential Equation)A : State matrix; B : input matrixy = Hx + Du (Output Equation)H : Output matrix; D : direct transmission matrix

**Example 1:** Consider the given series RLC circuit. Derive the differential equations that map the circuitry dynamics.



**Example 2:** Using the state-space concept, find the state-space model and analyze the transient dynamics of the series *RLC* circuit.

$$\frac{dv(t)}{dt} = \frac{1}{C}i$$
  
$$\frac{di}{dt} = \frac{1}{L}(-v_c - Ri + v(t))$$
  
$$x_1(t) = v_c(t); x_2(t) = i(t) \text{ (These are the states)}$$
  
$$v(t) \text{ is the control}$$

$$\frac{dx}{dt} = \begin{bmatrix} \frac{dx_1}{dt} \\ \frac{dx_2}{dt} \end{bmatrix} = \begin{bmatrix} \frac{dv_c}{dt} \\ \frac{di}{dt} \end{bmatrix} = \begin{bmatrix} 0 & \frac{1}{C} \\ -\frac{1}{L} & \frac{R}{L} \end{bmatrix} \begin{bmatrix} v_c \\ i \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} v_a = Ax + Bu$$

#### Continue with Values..

- Assume R = 2 ohm, L = 0.1 H, and C = 0.5 F, find the following coefficients.
- The initial conditions are assumed to be  $v_c(t_0)=v_{c0}=15$  V; and  $I(t_0) = i_0 = 5$  A.
- Let the voltage across the capacitor be the output;  $y(t) = v_c(t)$ . The output equation will be
- The expanded output equation in y
- The circuit response depends on the value of v (t)

$$A = \begin{bmatrix} 0 & 2 \\ -10 & -20 \end{bmatrix} \text{ and } B = \begin{bmatrix} 0 \\ 10 \end{bmatrix}$$
$$x_0 = \begin{bmatrix} x_{10} \\ x_{20} \end{bmatrix} = \begin{bmatrix} 15 \\ 5 \end{bmatrix}$$
$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} v_c \\ i \end{bmatrix} = Hx; H = \begin{bmatrix} 1 & 0 \end{bmatrix}$$

$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} v_c \\ i \end{bmatrix} + \begin{bmatrix} 0 \end{bmatrix} v_a = Hx + Du$$