

Broadband Optical Heterodyne Millimeter-Wave-over-Fiber Wireless Links Based on a Quantum Dash Dual-Wavelength DFB Laser

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Abstract—We demonstrate real-time broadband multi-Gb/s electrical RF synthesizer-free millimeter-wave (MMW) signals generation and wireless transmission at the 5G new radio (NR) frequency band of 47 GHz based on analog radio-over-fiber (A-RoF) fronthaul. This is enabled by a low noise, highly correlated, monolithic C-band semiconductor InAs/InP quantum-dash (QDash) dual-wavelength distributed feedback (DW-DFB) laser. One laser mode is encoded using 4-/6-Gbaud multilevel quadrature amplitude modulation (M-QAM) (16-/32-/64-QAM) baseband data signals, the other lasing mode is used as an optical local oscillator for optical-heterodyne remote up-conversion to a MMW carrier of 47.27 GHz. Consequently, optical baseband modulated data signals with data capacity up to 36 Gb/s (6-Gbaud \times 64-QAM) are transmitted through back-to-back (BtB) and 25-/50-km of standard single mode fiber (SSMF) before the MMW carrier is optically synthesized remotely for free space wireless data transmission and detection over up to 9-m. The end-to-end MMW-over-fiber (MMWoF) wireless link is thoroughly characterized exhibiting promising error-vector-magnitude (EVM) and bit-error-rate (BER) values. The 4-/6-Gbaud 16-QAM MMWoF wireless links achieve EVMs down to 6.32%/7.33%, 6.71%/7.78%, and 7.35%/8.91% through BtB, 25-km, and 50-km SSMF, respectively. Similarly, the EVM for

32-QAM and 64-QAM links is observed to be 5.56%/6.56% and 6.05%/6.62%, respectively. Moreover, in each case, the calculated BER is below the forward error correction (FEC) limit of 3.8×10^{-3} . The results corroborate the potential and viability of the QDash DW-DFB laser as a simple, efficient and cost-effective alternative to individual laser sources for deployment in broadband photonic MMWoF fronthaul systems of 5G wireless networks.

Index Terms—5G, broadband wireless communications, fronthaul, microwave photonics, millimeter-wave, optical heterodyning, quantum dash dual-wavelength DFB laser, radio-over-fiber.

I. INTRODUCTION

THE demand for broadband high speed, low latency, reliable and pervasive wireless connectivity has increased manifold since the inception of new technologies and bandwidth hungry applications, such as high definition video streaming, virtual and augmented reality (VR/AR), autonomous vehicles, artificial intelligence, and IoTs [1]–[3]. Furthermore, the Coronavirus crisis has heightened the urgency for broadband connectivity. It has been realized during the pandemic that broadband ubiquitous connectivity matters more than ever before and wireless technology is playing a crucial role. Not only helping people to stay connected around the globe but also enabling telehealth, remote learning and education, and especially remote work that keeps businesses, essential services and government operations running to meet daily needs [4]. This highlights the need and importance of 5G wireless networks, which promise to provide broadband high data rate wireless connectivity with high reliability and low latency pervasively.

The International Telecommunication Union (ITU) specifies 5G standards in International Mobile Telecommunications 2020 (IMT-2020) that identify three different usage scenarios including enhanced mobile broadband (eMBB), massive machine type communication (mMTC) and ultra-reliable low latency communication (URLLC). The eMBB features high speed broadband services with peak data download speed of more than 20-Gb/s and seamless user experience data rates of 100-Mb/s in wider coverage area with the expected speed of Gb/s in hotspot scenarios [5]. These ultra-high speed broadband and low latency services cannot be supported by the already depleted sub-6GHz RF spectrum. Therefore, 5G requires high

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frequency bands in the millimeter-wave (MMW) spectrum (30 GHz – 300 GHz) with plentiful available bandwidths for the realization of ultra-high speed and high-capacity broadband communications [6]. To this end, 3GPP has standardized higher frequency bands above 24.25 GHz for 5G new radio (NR) designated as frequency range 2 (FR2) (24.25 GHz – 52.6 GHz) [7] and work is underway for frequency bands beyond 52.6 GHz. Nonetheless, the generation and processing of high-speed and broadband MMW signals in an all-electrical setup is hindered by the bandwidth limitation of the electronic devices in addition to cost and complexity. Besides, the transmission of MMW signals over long distances is a real challenge. Consequently, broad bandwidth, simple, efficient, and cost-effective photonic millimeter-wave-over fiber (MMWoF) solutions are considered viable alternatives for MMW signal generation, processing, control and distribution in the optical domain for application in broadband wireless access networks [8], [9]. The optical devices and techniques that are used for MMW signals in conjunction with the bandwidth efficient analog radio-over-fiber (A-RoF) technology not only overcome the problem of high bandwidth requirements, transmission capacity and span limitation but also significantly reduces system complexity [1], [8]–[10], footprint, capital expenditure (CAPEX) and operating expenses (OPEX).

The basic notion of MMW signal generation and distribution in the optical domain through MMWoF links is based on the remote optical heterodyne beating of two phase-correlated optical signals having different wavelengths on a high speed photo detector (PD) after running over several tens of kilometers of optical fiber [11]. Photonic RF signal generation techniques can be used to generate RF signals of up to THz range, limited only by the frequency response of the PD. In this way, seamless fiber-wireless integration can be achieved [12] through generating RF MMW carriers optically, which is inevitable for 5G ultra-broadband wireless networks. Thus, optical heterodyne MMW over A-RoF can significantly simplify the overall fronthaul architecture compare to traditional architecture, especially the RF front end in the centralized cloud radio access network (C-RAN) environment of 5G. This is due to the fact that digital or analog data can be modulated onto either one or two optical signals in the baseband units (BBUs) located in the central office (CO) and the desired RF MMW carrier signal can be remotely synthesized optically right at the antenna eliminating the need for expensive components, such as digital to analog and analog to digital converters (DACs/ADCs) in the case of conventional digital RoF or electrical local oscillators and mixers in the case of conventional A-RoF

However, in the case of the optical heterodyne MMWoF configuration, the phase fluctuations of the two optical signals need to be highly correlated in order to ensure spectrally pure MMW carrier signal generation. Because if the two optical signals are not phase correlated, for instance in the case of two different free running lasers, then the resulting MMW signal ends up with high phase noise thus limiting the system performance [1], [13]. Conversely, if the phase noise of the two optical signals is correlated then the common noise of the two signals cancels out and a low phase noise MMW signal is generated. This entails

either the elimination or correlation of phase noise of the corresponding optical signals for the generation of spectrally pure and phase stable MMW carrier signals. Thus highly correlated and integrated optical sources are desirable for 5G MMWoF systems.

Several techniques and devices have been proposed and demonstrated for optical heterodyne MMW signal generation and data transmission in RoF links based on schemes such as using two individual single wavelength lasers [2], external modulation of a single mode laser [14]–[16], picking pairs of modes from optical frequency combs [13], [17], [18], sources based on colorless laser diode [19]–[22] and dual-wavelength optical sources [23], [24]. Similarly, several techniques have been reported for the phase locking and stabilization between the two optical heterodyne signals including optical injection locking [25], [26], optical-phase-locked-loop (OPLL) [27] and their combination [28].

These MMW sources and stabilization techniques offer better frequency correlation and stabilization than any two free-running lasers based schemes. Nevertheless, most of the MMW signal generation techniques utilizing phase locking and stabilization approaches rely on complex and expensive devices, such as modulators, microwave reference sources and other related components that add to the complexity and cost of the transmitter system. In addition, the bias drift associated with some of the aforementioned schemes involving external modulators can cause instability, hence limiting system performance. Consequently, besides attaining spectral purity and stability, the complexity and cost are the two key factors that need to be considered in the design of future 5G wireless systems

In recent years, semiconductor quantum dash (QDash) multi-wavelength lasers have gathered considerable attention for photonic MMW signal generation and applications in optical heterodyne MMWoF systems [1], [3], [29]–[41]. Compared with other technologies well-designed quantum confined nanostructures, such as semiconductor InAs/InP QDash lasers, have the distinctive advantage of achieving highly coherent and correlated optical signals with very low phase and intensity noise [29], [33], [38], [42]–[44], together with chip-scale integration viability. Moreover, for further spectral purity and phase stabilization in such highly integrated and correlated devices, a simple feedback mechanism of self-injection locking can be used [1], [30], [42], [45], [46] eliminating the need for comparatively more complex and expensive components, thus making them a suitable candidate as an optical MMW source for 5G MMWoF systems.

A number of InP based QDash multi-wavelength lasers are demonstrated as optical MMW sources in both C-band [1], [3], [29], [30], [33]–[35], [37]–[41] and L-band [31], [32], [35], [41]. These include some recent experimental demonstration of optical heterodyne MMWoF links using QDash passively mode-locked lasers (MLLs) or optical frequency combs [1], [29], [30], [37] in the C-band, QDash injection-locked laser diodes in the C-band [34], [35], [39]–[41] and L-band [31], [32], [36], [41], and two wavelengths in a free running state from a monolithically integrated QDash Y-coupled dual-DFB lasers source [38] and a QDash common-cavity dual-wavelength DFB

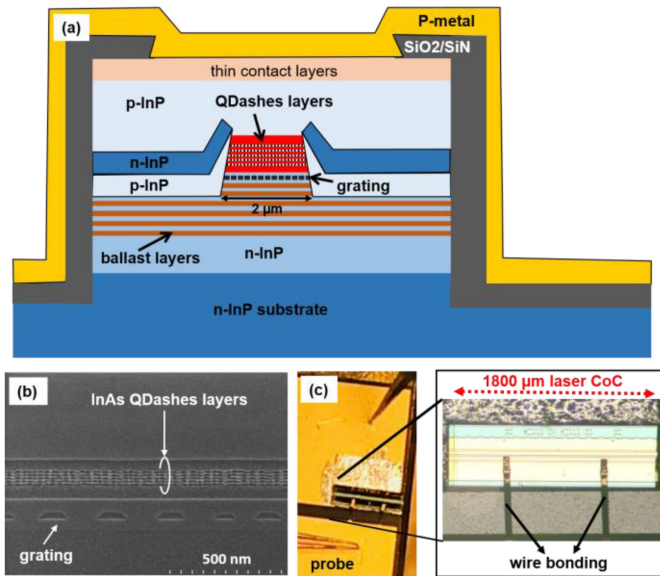


Fig. 1. (a) Schematic of the cross-section of the QDash DW-DFB laser (b) SEM image of the lateral cross-section through the middle of the mesa of the device showing synthesized aperiodic grating underneath 5 stack layers of QDashes (c) QDash DW-DFB laser CoC.

(DW-DFB) laser in the C-band [3]. Compared with other optical sources, such as QDash combs or QDash injection locked laser diodes and dual-mode devices based on separate DFB cavities or individual DFB lasers, QDash common-cavity DW-DFB laser offers a compact and simple solution for MMWoF systems by generating two different wavelengths from within the same cavity featuring high output optical power and low noise [33]. This results in high optical power per mode (significantly higher than for a QDash comb laser) and high spectral purity with low phase and relative intensity noise. Importantly, this makes the device comparatively simple and compact. Consequently, the QDash dual-wavelength DFB laser has the advantage of reducing system complexity and cost showing the potential to offer a simple and low cost solution for 5G optical heterodyne MMWoF systems.

In this paper, we extend our work on the use of an InAs/InP QDash DW-DFB laser in [3] for MMWoF systems with the experimental demonstration of broadband multi-Gb/s optical heterodyne MMWoF wireless links at 5G NR frequency of around 47.27 GHz. The demonstration includes electrical RF synthesizer-free real-time photonic generation, wireless transmission and detection of wide-bandwidth MMW M-QAM modulated (16-QAM, 32-QAM and 64-QAM) data signals having symbol rate of 4- and 6-GHz with bit rates ranging from 16 Gb/s (4-GBaud \times 16-QAM) to 36 Gb/s (6-GBaud \times 64-QAM) over hybrid fiber-wireless links comprising of back-to-back (BtB), 25-km and 50-km SSMF and 2-m to 9-m free-space wireless channel. After long-reach transmission of baseband data modulated optical signal in 25-/50-km SSMF, the MMW carrier is optically-synthesized remotely through optical-heterodyning for free space wireless data transmission. This ensures the transmission and distribution of ultra-high frequency MMW signals over long distances with better performance. Moreover,

a thorough end-to-end optical-MMW link analysis is performed with different M-QAM, fiber spans, MMW link distances and data transmission bandwidths to evaluate the transmission performance of the proposed QDash DW-DFB laser based MMWoF system in terms of EVM, BER, constellations and eye diagrams. To the best of our knowledge, this is the first experimental demonstration to realize MMWoF wireless links at the potential 5G NR frequency band of around 47.27 GHz with real-time wireless transmission and detection of wideband MMW M-QAM (16-/32-/64-QAM) modulated data signals having a maximum bit rate of 36 Gb/s using a free-running InAs/InP QDash DW-DFB laser. The generated MMW frequency range from 46 GHz to 48 GHz of the device falls within the potential 3GPP 5G NR standard band (n262) of frequency range 2 (FR2) offering a promising optical MMW source for 5G MMWoF systems.

The remainder of this paper is organized as follows. Section II presents a brief description of the design, implementation and characterization of the QDash DW-DFB laser. Section III presents the proposed QDash DW-DFB laser based MMWoF wireless system design and detailed implementation. It is followed by the experimental results and discussion in Section IV. Finally, section V presents the conclusion.

II. QUANTUM DASH DUAL-WAVELENGTH DFB LASER

The laser source used in this study is an InP based p-n-blocked buried heterostructure common cavity dual-mode QDash DFB laser. Fig. 1(a) shows a schematic of the cross-section of the laser. The gain region of the device consists of 5 layers of InAs QDashes in a 170 nm thick InGaAsP waveguiding core with 10 nm $\text{In}_{0.81}\text{Ga}_{0.184}\text{As}_{0.392}\text{P}_{0.608}$ (1.15Q) barriers. This active region is surrounded by n-type and p-type InP cladding layers as shown in Fig. 1(a). In the lower n-type InP cladding, 1.03Q ballast layers are employed to pull the mode into the n-type region, which helps in reducing cavity loss and increase efficiency. This also reduces the spontaneous emission coupling to the optical mode. The InAs QDash material was grown on (001)-oriented n-type InP substrate using chemical beam epitaxy (CBE) [47] followed by further etch and regrowth steps using metal-organic chemical vapor deposition (MOCVD) to create buried the heterostructure. The 1800 μm long laser waveguide was fabricated through standard photolithograph with dry- and wet-etching and contact metallization techniques.

An e-beam written novel synthesized aperiodic non-uniform diffraction grating is incorporated below the QDash active layers in the n-type InP cladding as shown in Fig. 1(a). This can also be seen in the scanning electron microscope (SEM) image of the lateral cross-section of the device below the 5 stack layers of QDashes in Fig. 1(b). This generates two highly correlated longitudinal modes simultaneously at different wavelengths with a drive current and temperature controlled tunable MMW frequency range between 46 and 48 GHz from within the single optical cavity of the device with QDash active layers. A full description of the grating can be found in [33]. After the growth of the laser core, a 2 μm wide mesa is formed by etching through the 1.15Q waveguide core and grating layer using dielectric mask followed by selective area overgrowth of pnp blocking

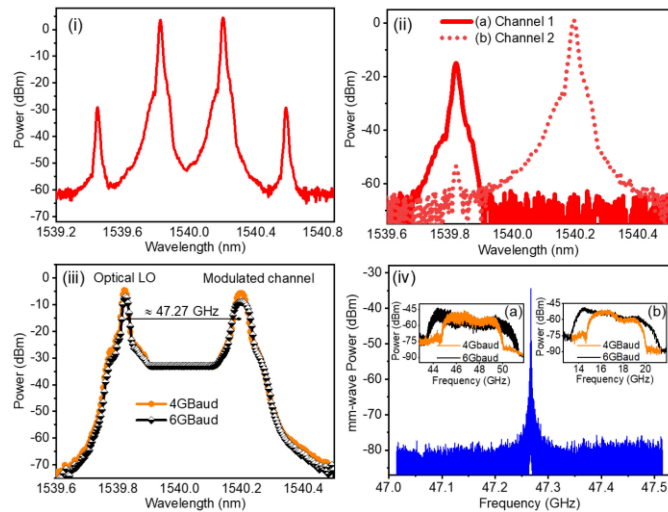


Fig. 2. Measured (i) optical spectrum of the QDash DW-DFB laser at the BBU after isolator (ii) optical spectrum of (a) channel 1 (optical LO) and (b) channel 2 (data channel) after TOBPF1 and TOBPF2, respectively, (iii) unmodulated (optical LO) and optical modulated data channel at the PD, and (iv) typical spectrum of the MMW carrier at the UE before down-conversion with insets showing the received 4-Gbaud and 6-Gbaud modulated 16-QAM data signals occupying 5.4 GHz and 8.1 GHz transmission bandwidth, respectively (a) before mixer and (b) after mixer by down-converting to IF.

layers to confine the carriers to the waveguide mesa. Finally the dielectric mask was removed and p-type InP cladding and contact layers were grown. After cleaving both facets of the device were AR coated.

The 1800 μm long laser chip is mounted on a commercially available Aluminium Nitride (AlN) carrier to provide mechanical support to the device. The chip on carrier (CoC) is shown in Fig. 1(c). The QDash laser chip is bonded to the carrier with Gold Tin (AuSn) providing the cathode connection, with the top contact being wire bonded for the anode connection as shown in Fig. 1(c). The laser threshold current is around 70 mA and the device can provide up to 50 mW average output power at higher injection current.

The output optical spectrum of the laser is comprised of two dominant optical modes with equal amplitudes at two different wavelengths (1539.821 nm and 1540.195 nm) around 0.374 nm apart as shown in Fig. 2(i). Generating both frequencies in the same cavity significantly reduces noise and linewidth of the generated light. Two four wave mixing (FWM) modes can also be seen in the spectrum that originate from the self-mixing of the two primary optical modes within the laser cavity [33]. This shows nonlinear effects of the cavity, which is believed to result in phase locking of the modes, hence reducing their relative amplitude and phase variation. This further stabilizes modal amplitudes and relative phase of the QDash DW-DFB laser. The integrated relative intensity noise (RIN) and optical linewidth for each individual channel are measured to be typically less than -150 dB/Hz in the frequency range from 10 MHz to 20 GHz and 30 kHz with the lowest recorded down to -158.3 dB/Hz and 15.83 kHz, respectively, in free running operation. This subsequently results in generating a low phase noise optical

heterodyne MMW carrier beat signal with a 3-dB linewidth of typically 40 kHz or narrower [3], [33].

Note that by changing drive current and temperature the mode spacing can be tuned by around 0.0158 nm, which corresponds to tuning MMW frequency range from 46 GHz to 48 GHz. However, the RF beat note of the QDash DW-DFB laser can be adjusted from the GHz to the THz range by modifying the design of the synthesized aperiodic grating of the device for the desired spectrum.

III. DESIGN AND EXPERIMENTAL CONFIGURATION OF QDASH DW-DFB LASER BASED OPTICAL HETERODYNE MMWOF WIRELESS SYSTEM

A schematic of the experimental configuration for the proposed optical heterodyne MMWof wireless system is depicted in Fig. 3. In our demonstration, we emulate a typical 5G C-RAN fronthaul architecture where the experimental setup is comprised of a baseband unit (BBU) connected to an RF electrical LO-free optical heterodyne synthesizer based remote radio unit (RRU) through a 25-km or 50-km SSMF link followed by 2-m to 9-m MMW free space indoor RF wireless link connecting to the MMW user equipment (EU). Fig. 4 shows a detailed component pictorial view of the experimental setup including the MMWof wireless transmission link. In the BBU, the InAs/InP QDash common cavity DW-DFB laser is employed as the main optical source. In our experiments the CoC is placed on a copper block with a thermoelectric cooler (TEC) underneath for temperature control. The temperature of the CoC and the laser bias current are controlled through a laser diode controller (ILX Lightwave, Model LDC-3722) where the laser is biased through a pair of DC electrical probes as shown in Fig. 3 and the inset of Fig. 4. Throughout our experiments, the temperature and injection current of the CoC are maintained at 18°C and 360 mA, respectively.

In the BBU, the laser output power is coupled from its front facet through a lensed single mode polarization maintaining fiber (SMPMF) followed by a two stage PM isolator to avoid back reflection into the laser cavity. The light is then split into two paths by employing a 10/90 PM optical coupler (OC_1) followed by two tunable optical band pass filters (TOBPFs). The two optical filters, TOBPF_1 and TOBPF_2 , separate the two optical modes having a frequency spacing of around 47.27 GHz into channel 1 (1539.821 nm) and channel 2 (1540.195 nm) as shown in Fig. 2 (ii). Channel 2 is used as the optical modulated channel for long-reach data transmission whereas channel 1 is used as an un-modulated optical LO for remote optical heterodyning to optically synthesize the MMW carrier for free space wireless data transmission.

As a proof of concept, a QAM optical transmitter having thermally stable in-phase/quadrature (I/Q) Lithium Niobate (LiNbO_3) Mach-Zehnder modulator (MZM) and linear data driver electrical amplifiers (EAs) is used in our experiments to realize the data modulated optical channel. Consequently, channel 2 is modulated with baseband 16-, 32-, and 64-QAM data signals having symbol rate of 4-GHz and 6-GHz that are

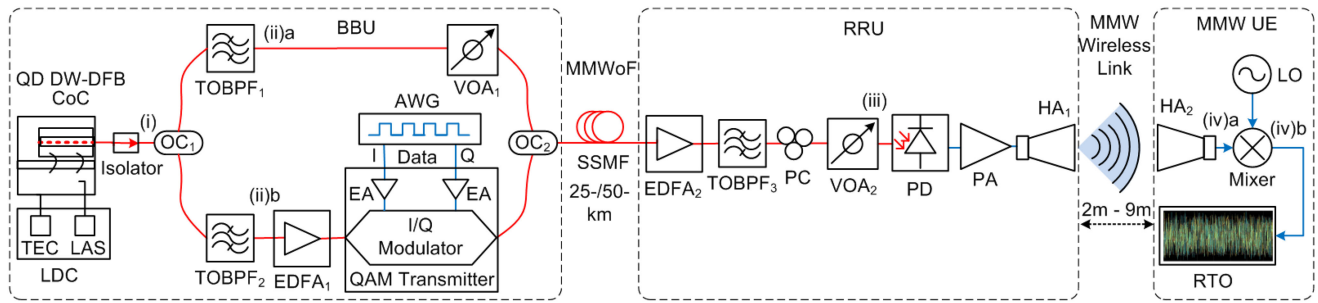


Fig. 3. Schematic of the system experimental setup for the optical heterodyne MMWoF wireless links based on the QDash DW-DFB laser. BBU: baseband unit; RRU: remote antenna unit; MMW UE: millimeter-wave user equipment; CoC: chip-on carrier; OC: optical coupler; TEC: thermoelectric cooler; LAS: laser; LDC: laser diode controller; TOBPF: tunable optical band pass filter; EDFA: erbium doped fiber amplifier; AWG: arbitrary waveform generator; EA: electrical amplifier; VOA: variable optical attenuator; PC: polarization controller; PA: power amplifier; HA: horn antenna; LO: local oscillator; RTO: real-time oscilloscope.

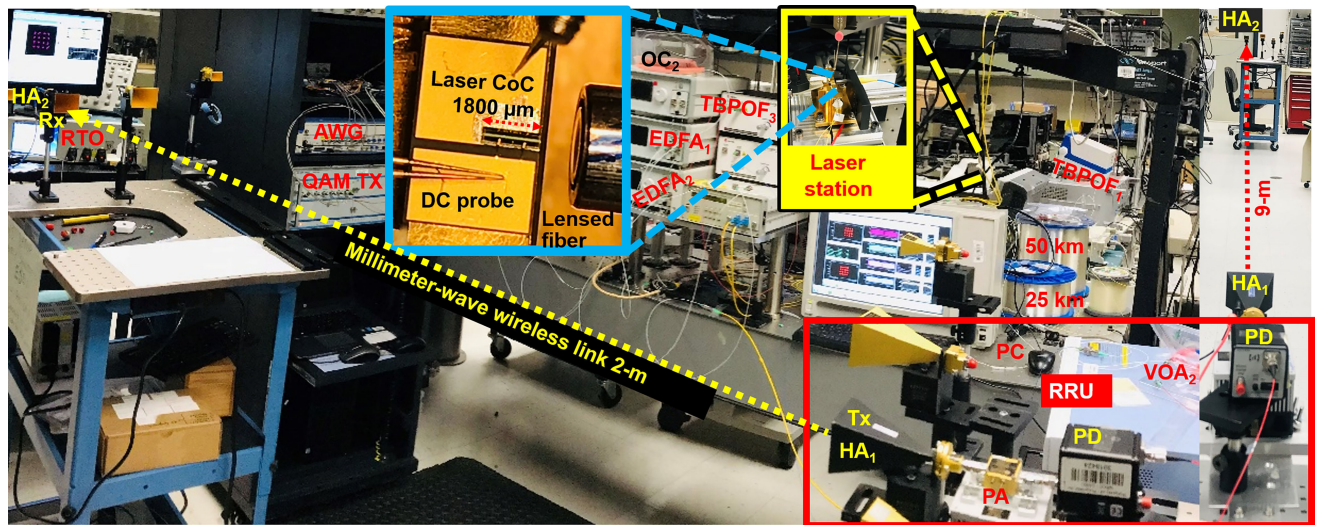


Fig. 4. Photos of the Laboratory experimental setup of QDash DW-DFB laser based optical heterodyne MMWoF system with insets showing the QDash DW-DFB laser CoC on a laser testing station, MMW wireless transmission links (2-m to 9-m), SSMFs (25 & 50-km), wireless transmitter (Tx) and receiver (Rx) and all the other key components of BBU, RRU and MMW UE.

generated electronically by a 65-GSa/s arbitrary waveform generator (AWG) (Keysight M8195A) with a pseudo-random binary sequence (PRBS) pattern of $2^{11}-1$ bits where the bit sequence is mapped onto the I and Q components of the signals. The corresponding data signals are then passed through a root raised cosine (RRC) filter with a roll-off factor of 0.35 for Nyquist pulse shaping. After pulse shaping, the signals are resampled and channel corrections are applied for amplitude flatness before feeding them to the optical transmitter. The data signals are then amplified by two linear data EAs in the optical transmitter before they are fed to I/Q MZM for modulation. Before modulation, the data channel is boosted with an erbium-doped fiber amplifier (EDFA₁) to compensate for the insertion loss of the filter and optical modulation transmitter. A PM variable optical attenuator (VOA₁) is also employed in the un-modulated path to equalize its power level to that of modulated channel since the two paths have different losses.

The effective path length difference between the two arms of the transmitter can affect the degree of phase correlation between the two wavelength signals, potentially leading to relatively high phase noise in the generated mm-wave carrier [1], [13].

This decorrelation between the two channels can be avoided by incorporating an optical delay line in the unmodulated path to equalize the effective path lengths. In this way the QAM modulated 47.27 MMW carrier signal is realized in the optical domain with a maximum data rate of 36 Gb/s. Finally, the modulated and unmodulated optical channels are combined in a 50/50 OC₂ and transmitted over the optical fiber link to the RRU. The optical fiber link is comprised of 25-km or 50-km SSMF spool as highlighted in Fig. 4.

The received optical signal at the RRU is amplified through EDFA₂ followed by OBPF₃ to remove the amplified spontaneous emission (ASE) noise as shown in Fig. 3 and Fig. 4. The optical spectrum of the corresponding received modulated and un-modulated carriers before the PD is shown in Fig. 2 (iii). The photo-mixing output of these two optical frequency signals on a high speed PD (Newport Model-1014), is directly attached to a 22-24 dBi Horn antenna (HA₁) (WR-22) through an RF Q-band power amplifier (PA) with a nominal gain of 45 dB. Thus, a MMW carrier of around 47.27 GHz is optically synthesized remotely without using any electrical LO and the optical data is translated to the corresponding MMW carrier. The data signal on

the 47.27 GHz MMW carrier is subsequently transmitted over the 2-m to 9-m wireless link. A polarization controller (PC) is incorporated in the RRU to adjust the polarization direction of the incoming optical signal before photo-mixing to ensure maximum output power at the output of PD. A PM VOA₂ is also employed at the RRU before the PD to control the received incident power on the PD for different wireless link measurements. Note that similar to the transmitter system at the BBU, we use off-the-shelf components at the RRU to demonstrate our proof-of-concept broadband multi-Gb/s MMWoF system. Nevertheless, all of these components, where the EDFAs can be replaced by semiconductor optical amplifiers (SOAs), could be heterogeneously integrated to microwave photonic circuits making the BBU and the RRU more simple compared with the conventional RoF fronthaul systems.

The MMW data signal is received by another identical Horn antenna at the user equipment (UE) over the free space wireless link. This MMW signal is then down-converted to an intermediate frequency (IF) of around 17.3 GHz with an electrical LO before capturing into a real-time oscilloscope (RTO) for processing as shown in Fig. 3. Fig. 2 (iv) shows the corresponding received MMW carrier of 47.27 GHz along with the 4-Gbaud and 6-Gbaud modulated 16 QAM data signals in the insets (a) before and (b) after down-conversion, occupying transmission bandwidth of 5.4 GHz and 8.1 GHz, respectively. After down-converting the received MMW signal to an IF, as shown in the inset (b) of Fig. 2 (iv), it is coherently detected and processed in real-time by employing vector signal analysis software (SignalVu) on a 100 GSa/s Tektronix DPO73304SX oscilloscope having 33-GHz analog bandwidth. The signal undergoes several digital signal processing (DSP) steps including resampling, clock recovery, digital down-conversion, match filtering, synchronization, and adaptive equalization before it is demodulated to measure the error vector magnitude (EVM) and calculate the bit error rate (BER). For match filtering, an RRC matched filter with roll-off factor of 0.35 is employed to recover the baseband IQ data and to minimize inter-symbol interference (ISI). Similarly, a decision-directed, feed-forward (FIR) adaptive equalizer is used to compensate for linear distortions. Finally, the system integrity and communication performance is evaluated in terms of measured EVM and calculated BER [48].

IV. EXPERIMENTAL RESULTS AND DISCUSSION

The performance of the whole MMWoF wireless system is thoroughly analyzed in terms of EVM, BER, constellations, and eye diagrams. Multilevel QAM signals including 16-QAM, 32-QAM and 64-QAM are used to realize MMW wireless links with link capacities ranging from 16-Gb/s to 36-Gb/s. The system performance is then evaluated for different line rates over a fixed MMW wireless link through different fiber lengths by varying the received optical power (ROP), and for different wireless link lengths by changing the wireless transmission distance at a fixed ROP through a 25 km SSMF.

Fig. 5(a) and (b) summarize the EVM performance results of 4-Gbaud and 6-Gbaud 16-QAM signals with data rates of 16 Gb/s and 24 Gb/s, respectively, over a 2-m MMW wireless link

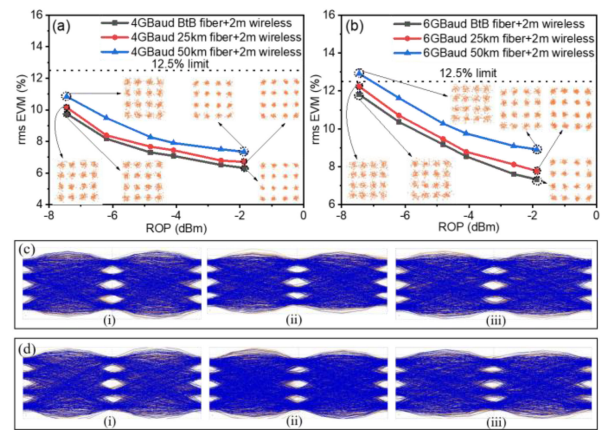


Fig. 5. rms EVM measured at the UE versus ROP at the PD for 16-QAM modulated data signals with (a) 4-Gbaud symbol rate occupying 5.4 GHz bandwidth and (b) 6-Gbaud symbol rate occupying 8.1 GHz bandwidth. Eye diagrams recorded at the UE over 2-m wireless link at the ROP of ~ -1.9 dBm for 16-QAM (c) 4-Gbaud and (d) 6-Gbaud through (i) BtB, (ii) 25-km, and (iii) 50-km SSMF link, respectively.

through BtB, 25-km and 50 km SSMF. The different fiber lengths and signal bandwidths are employed to analyze and compare the fiber and signal bandwidth induced impact on the performance with BtB as a reference. In each case, the root mean square (rms) EVM of the 4-Gbaud and 6-Gbaud 16-QAM wirelessly received MMW signals is measured in real-time at the MMW UE shown in Fig. 3 for different ROPs at the PD. Note that in order to analyze the effect of the optical fiber transmission on the wireless link, the ROP for all of the three fiber lengths is kept identical at the PD by using VAO₂ as shown in Fig. 3.

The results show that for all 16-QAM signals, the EVM is below the 3GPP standard requirement of 12.5%, except for the case of 6-Gbaud transmission through a SSMF length of 50 km, where it is slightly over the limit at the ROP of around -7.5 dBm. The inset of Fig. 5(a) and (b) also include clear constellation diagrams showing the successful detection of the corresponding 4-Gbaud and 6-Gbaud 16-QAM MMW signals at the ROP of -1.9 dBm and -7.5 dBm, respectively. This is also evident from the open eye diagrams of the corresponding wirelessly received 4-Gbaud and 6-Gbaud 16-QAM signals as depicted in Fig. 5(c) and (d), respectively, at the ROP of -1.9 dBm for the (i) BtB, (ii) 25 km, and (iii) 50 km scenarios. Moreover, Fig. 6 depicts the rms EVM as function of the baud rate for the successful detection of 16-QAM modulated optical heterodyne MMW signals, centered at around 47.27 GHz, over 2-m wireless link through BtB, 25-km and 50-km SSMF at the ROP of -1.9 dBm.

Fig. 7 plots the calculated BER versus the ROP for both 4-Gbaud and 6-Gbaud 16-QAM MMW signals through BtB, 25 km and 50 km SSMF. The BER is calculated based on the measured rms EVM using the relationship between EVM and BER derived in [48]. It is noted that the BER increases with the increase in baud rate from 4-Gbaud to 6-Gbaud at the same ROP. A BER $< 10^{-9}$ is observed for 4-Gbaud transmissions at the ROP of > -2.5 dBm for all of the three scenarios. However, for 6-Gbaud transmissions, the same is only observed in the case of BtB scenario as shown in Fig. 7. Overall, the

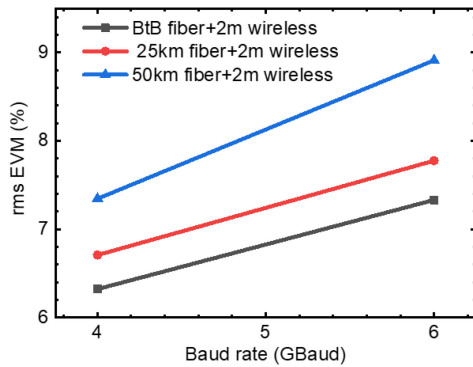


Fig. 6. rms EVM as a function of the baud rate for 16-QAM MMWoF wireless links at the ROP of ~ -1.9 dBm.

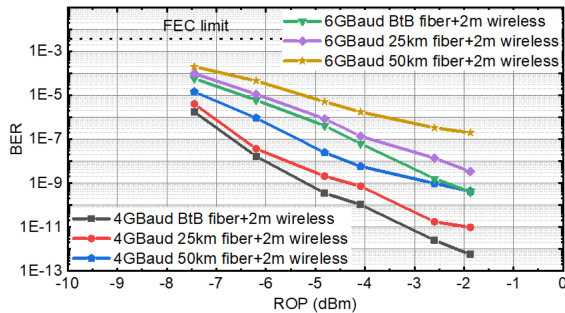


Fig. 7. Calculated BER as a function of ROP for 4-GBaud and 6-GBaud 16-QAM data signals measured over 2-m MMW wireless link through BtB, 25-km and 50-km SSMF.

BER performance for both 4-GBaud and 6-GBaud transmissions under all scenarios is below the standard FEC limit of 3.8×10^{-5} .

Generally, an increase in performance degradation is observed with an increase in bandwidth of the transmitted signals from 5.4 GHz to 8.1 GHz having baud rate of 4-GBaud and 6-GBaud, respectively, as can be seen from Fig. 6. This is attributed to the fact that the MMWoF system is more prone to noise and propagation loss at wideband signal operation. However, it is observed that on average, the fiber length of 25 km induces slight performance degradation with EVM penalty of around 0.4 dB for both 4-GBaud and 6-GBaud optical heterodyne MMW wireless transmissions with respect to their BtB scenarios. This degradation further increases to around 1.1 dB and 1.2 dB for 4-GBaud and 6-GBaud transmissions, respectively, in the case of 50-km fiber. Similarly, an average penalty of around 0.7 dB and 0.8 dB is observed between 25-km and 50-km transmissions of 4-GBaud and 6-GBaud mm-wave signals, respectively. Nevertheless, an average degradation of about 1.7 dB is observed between the transmissions of 4-GBaud and 6-GBaud 16-QAM MMW signals through BtB, 25-km and 50-km SSMF, which corresponds to an average optical power penalty of around 2.8 dB. This is ascribed to the very nature of the wide bandwidth operation encountering relatively more noise, fiber dispersion and channel propagation loss at the MMW frequency. A typical spectrum of the received modulated MMW data signals at the UE with 5.4 GHz and 8.1 GHz transmission bandwidths both before and after down-conversion can be seen in the inset (a)

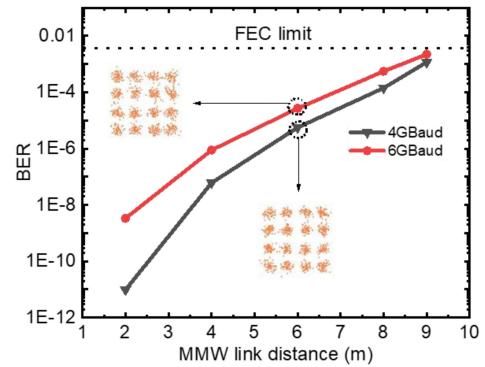


Fig. 8. BER versus MMW wireless link distance at ROP of ~ -1.9 dBm for 4-GBaud and 6-GBaud 16-QAM received signals through 25-km SSMF.

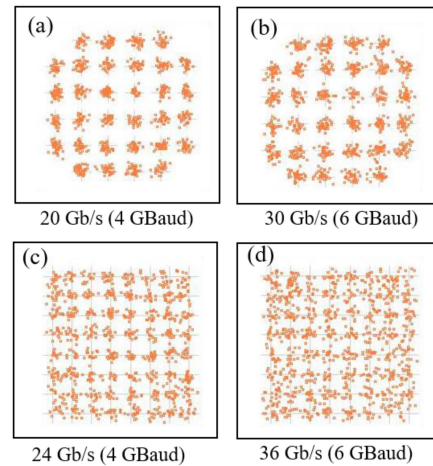


Fig. 9. Wirelessly received signal constellations of (a) 4-GBaud 32-QAM (20 Gb/s), (b) 6-GBaud 32-QAM (30 Gb/s), (c) 4-GBaud 64-QAM (24 Gb/s), and (d) 6-GBaud 64-QAM (36 Gb/s).

and (b) of Fig. 2 (iv), respectively, for the ROP of around -1.9 dBm at the PD. The uneven envelop of the received modulated signals towards the high frequency end is due to the operating frequency range of the PD, PA and antennas.

Thus, the EVM and BER performance degradation with fiber length in the aforementioned cases is believed to be due to the accumulated dispersion encountered over the optical fiber link along with the effective path length difference between the two optical carriers at the transmitter [1] and the corresponding wireless propagation path loss. To further improve the results, an optical delay line can be incorporated in the un-modulated path at the transmitter side and dispersion compensation can be employed to further reduce the effect of optical carrier decorrelation at the PD [13]. Besides, it is observed that polarization misalignment impacts the results, which can be avoided by employing proper polarization control. In addition, based on the antennas configuration in our experiments, accurate direct line-of-sight (LOS) path is required from the transmitter antenna to the receiver antenna to establish a link with better performance.

The effect of the MMW wireless link length on the performance of the system is also investigated. Fig. 8 shows the BER performance at different wireless link distances through a 25-km SSMF at a fixed ROP of around -1.9 dBm for both

4-GBaud and 6-GBaud 16-QAM signals. Their corresponding constellation diagrams for a wireless link distance of 6-m are also shown in Fig. 8. The measured EVM values for these constellation diagrams of 4-GBaud and 6-GBaud received signals were 10.33% and 11.26% respectively. The results show that BER performance of under the FEC limit of 3.8×10^{-3} is achieved for both 4- and 6-GBaud MMW signals wireless transmissions with a maximum MMW link distance of 9-m at the ROP of -1.9 dBm. It is apparent that the 4-GBaud 16-QAM signal achieves better BER performance as compared to 6-GBaud 16-QAM. However, in both cases, the performance deteriorates as the wireless distance increases. In general, it is observed that increase in the RF link distance or decrease in the ROP degrades the system performance. This is attributed to the wireless propagation path loss along with imprecise antenna alignment and increase in noise level at the received low MMW signal power that results in low electrical signal to noise ratio (ESNR) contributing to the performance degradation. Moreover, since we use directional antennas in our experiments their alignment becomes critical as the wireless distance increases and the lack of proper LOS impacts the link performance.

Finally, Fig. 9(a) and (b) and Fig. 9(c) and (d) show the constellation diagrams for 4- and 6-GBaud 32-QAM and 64-QAM MMWoF wireless links, respectively, with 25-km SSMF and 2-m wireless distance at the ROP of around -1.9 dBm at the PD. The rms EVM values for 4-/6-GBaud 32-QAM and 64-QAM links with a transmission bandwidth of 5.4-/8.1-GHz were measured to be 5.56%/6.56% and 6.05%/6.62%, respectively. Fig. 9(a) and (b) show the comparison between 4-GBaud and 6-GBaud 32-QAM MMW signals transmissions with the calculated BERs of 7.26×10^{-9} and 7.06×10^{-7} , respectively. Similarly, Fig. 9(c) and (d) show the comparison between the MMWoF wireless transmission of 4-GBaud and 6-GBaud 64-QAM MMW signals with the calculated BER of 8.94×10^{-5} and 2.84×10^{-4} , respectively. This constitutes a link data throughput of 20 Gb/s and 30 Gb/s in the case of 4- and 6-GBaud 32-QAM signals, and 24 Gb/s and 36 Gb/s in the case 64-QAM signals, respectively.

The results show that the QDash DW-DFB laser based optical heterodyne MMWoF wireless links achieve EVM and BER performance of under the FEC limit in all cases for both 4- and 6-GBaud wideband 47.27 GHz MMW wireless signals transmissions with a minimum BER of 5.7×10^{-13} and 4×10^{-10} , respectively, and a maximum bit rate of 36 Gb/s. These results demonstrate the capabilities of QDash DW-DFB lasers with the MMW frequency range within the 5G NR standard band (n262) of FR2 as a promising optical MMW source for 5G optical heterodyne synthesizer based MMWoF wireless systems.

V. CONCLUSION

We successfully demonstrate various real-time broadband optical heterodyne synthesizer based MMWoF wireless links featuring multi-Gb/s data rates with a maximum data capacity of 36 Gb/s (64-QAM \times 6-GBaud) having EVM and BER below the standard 7% overhead FEC limit of 3.8×10^{-3} using a free running QDash DW-DFB laser. The results indicate that the optical heterodyning of two highly correlated and low noise optical

signals is an efficient method for transporting and distributing high frequency MMW data signals in the optical domain over long distances. After long-reach transmission of these basebands data modulated optical signals, the corresponding desired MMW modulated carrier signal can be optically-synthesized remotely through photonic up-conversion in the MMW RoF systems for free space wireless data transmission. This greatly simplifies the overall system by preserving any modulated data format on the optical signals and translating it to the corresponding MMW carrier signal. This eliminates front end expensive components, such as electrical LOs and ADCs/DACs at the RRU. This highlights the potential of low noise, correlated and monolithically integrated InAs/InP QDash DW-DFB lasers for deployment as an alternative to individual DFB lasers in 5G optical heterodyne MMWoF systems.

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