



# ELG4177 - DIGITAL SIGNAL PROCESSING Tutorial 6

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# Midterm Part2 Solution

## Introduction to Lab#6

1. [4 marks] Consider an unknown LTI system. With an input  $x[n] = \delta[n]$ , the system's output was found to be:  $y[n] = 3\delta[n-1] - \delta[n-2] + 7\delta[n-4] + 2\delta[n-6]$ .

What is the system's output if the input is given by  $x[n] = 3\delta[n] - 4\delta[n-1]$ .

Q1: the system is LTI.

$$x[n] = 3\delta[n] - 4\delta[n-1]$$

$$\text{when } x_1[n] = \delta[n], \quad y_1[n] = 3\delta[n-1] - \delta[n-2] + 7\delta[n-4] + 2\delta[n-6]$$

when  $x_2[n] = \delta[n-1]$ ,

$$y_2[n] = y_1[n-1] = 3\delta[n-2] - \delta[n-3] + 7\delta[n-5] + 2\delta[n-7]$$

$$\therefore y[n] = 3y_1[n] - 4y_2[n]$$

$$= 9\delta[n-1] - 3\delta[n-2] + 21\delta[n-4] + 6\delta[n-6] - 12\delta[n-2] + 4\delta[n-3] - 28\delta[n-5] - 8\delta[n-7]$$

$$= 9\delta[n-1] - 15\delta[n-2] + 4\delta[n-3] + 21\delta[n-4] - 28\delta[n-5] + 6\delta[n-6] - 8\delta[n-7].$$



Q1:- The system is LTI

$$x[n] = \delta[n]$$

$$y[n] = h[n] = 3\delta[n-1] - \delta[n-2] + 7\delta[n-4] + 2\delta[n-6]$$

$$y[n] = x[n] \otimes h[n]$$

$$\delta[n] \otimes h[n] = h[n]$$

$$H(e^{j\omega}) = 3e^{-j\omega} - e^{-2j\omega} + 7e^{-4j\omega} + 2e^{-6j\omega}$$

$$Y(e^{j\omega}) = H(e^{j\omega}) X(e^{j\omega})$$

$$= 9e^{-j\omega} - 12e^{-2j\omega} - 3e^{-3j\omega} + 4e^{-3j\omega} + 21e^{-4j\omega} - 28e^{-5j\omega} + 6e^{-6j\omega} - 8e^{-7j\omega}$$

$$y[n] = 9\delta[n-1] - 15\delta[n-2] + 4\delta[n-3] + 21\delta[n-4] - 28\delta[n-5] + 6\delta[n-6] - 8\delta[n-7]$$



2. [8 marks] Consider the discrete-time system described by:  $y[n] = nx[n]$ , where  $x[n] = \delta[n - n_0]$ . Use two different methods to calculate the Fourier transform of  $y[n]$ ,  $Y(e^{j\omega})$ .

Q2:  $y[n] = n\delta[n - n_0]$

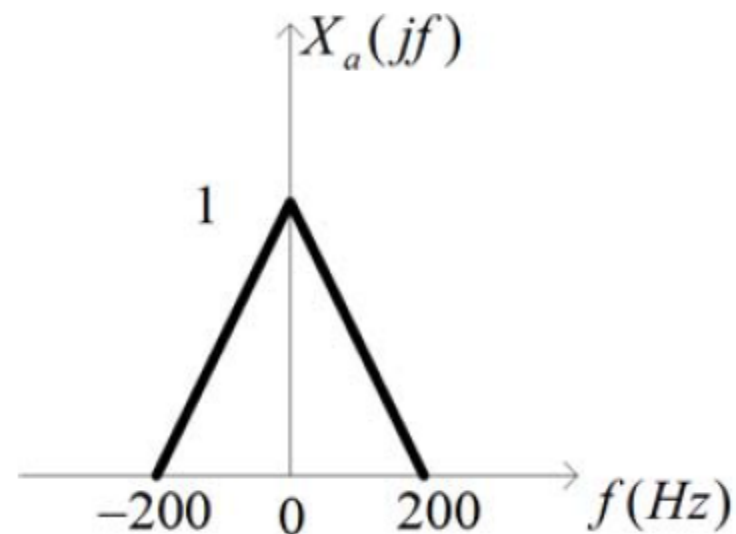
$$(1) Y(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} n\delta[n - n_0] e^{-j\omega n} = n_0 e^{-j\omega n_0}$$

$$(2) \delta[n - n_0] \leftrightarrow e^{-j\omega n_0}$$

$$n\delta[n - n_0] \leftrightarrow j \frac{d}{d\omega} (e^{-j\omega n_0})$$

$$= j \cdot e^{-j\omega n_0} \cdot (-jn_0) = n_0 e^{-j\omega n_0}$$

3. [10 marks in total] Suppose that  $x_a(t)$  is an analog signal whose spectrum is shown by the following figure. Sample  $x_a(t)$  in the time domain using the Nyquist sampling period ( $T$ ) and obtain  $x[n] = x_a(nT)$ .



- a. [4 marks] Sketch  $X(e^{j\omega})$  which is the Fourier Transform of  $x[n]$ .
- b. [6 marks] Increase the sampling rate by 2. Illustrate in the frequency domain the process of upsampling. In the illustration, draw the Fourier transform  $X_e(e^{j\omega})$  of the expanded signal  $x_e[n]$ , the frequency response of the low-pass filter  $H_i(e^{j\omega})$ , and the Fourier transform  $X_i(e^{j\omega})$  of the upsampled signal  $x_i[n]$ .

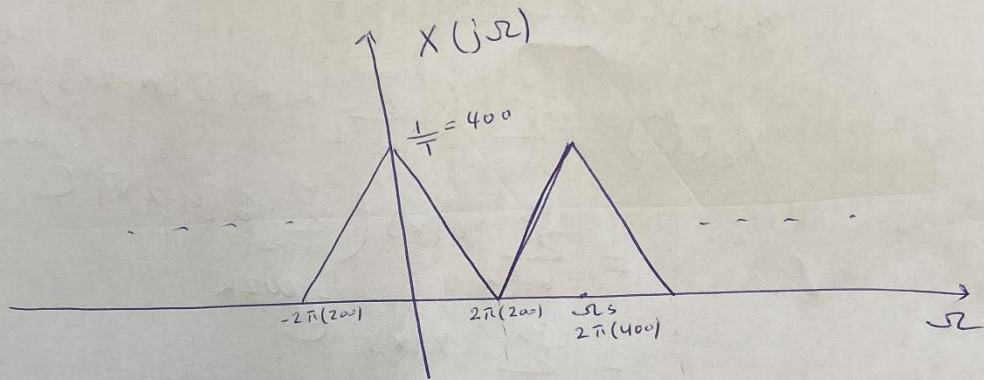


Q3 :-

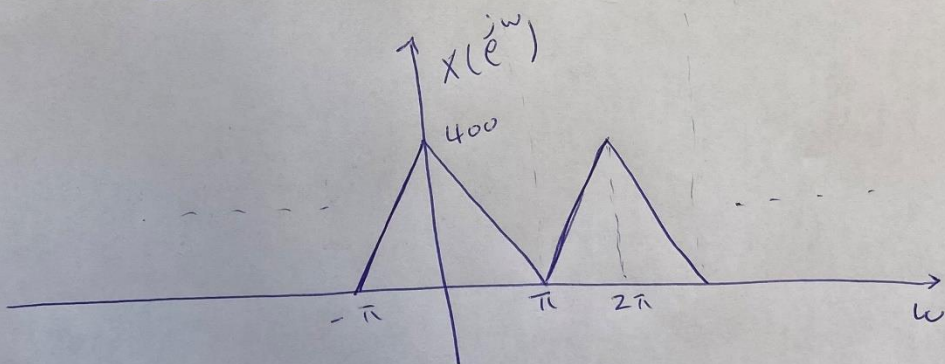
(a) Nyquist Sampling

$$f_0 = 200 \text{ Hz} \Rightarrow f_s = 2 f_0 = 400 \text{ Hz}$$

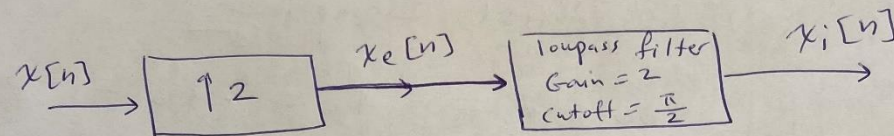
$$T = \frac{1}{400}$$



$$\Omega T = \omega$$



(b)



$$x_e[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - Lk], \quad L=2$$

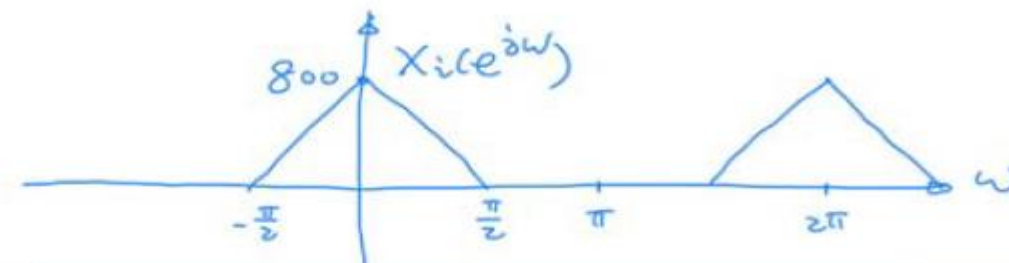
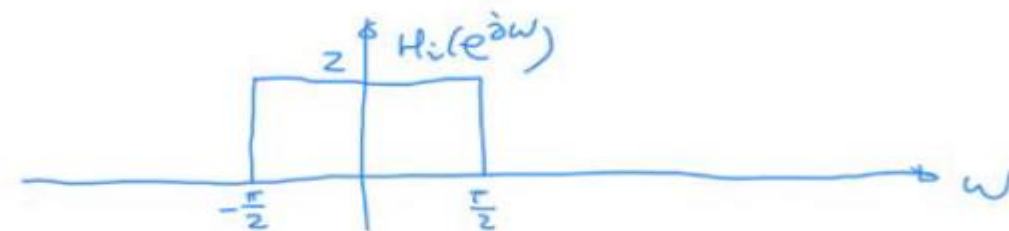
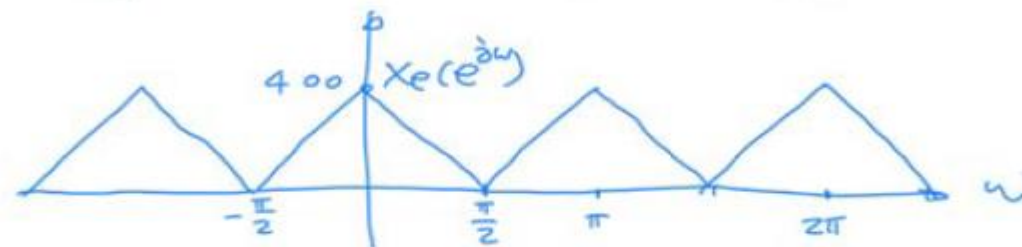
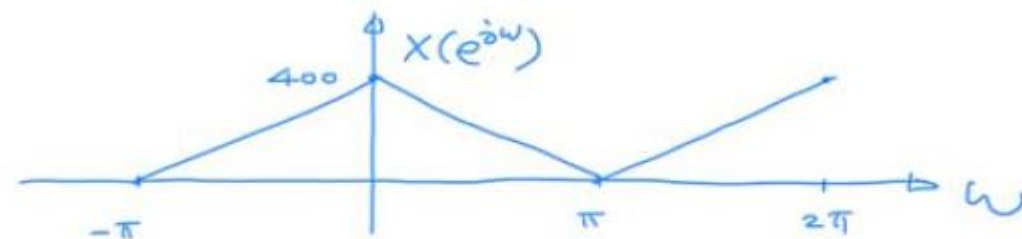
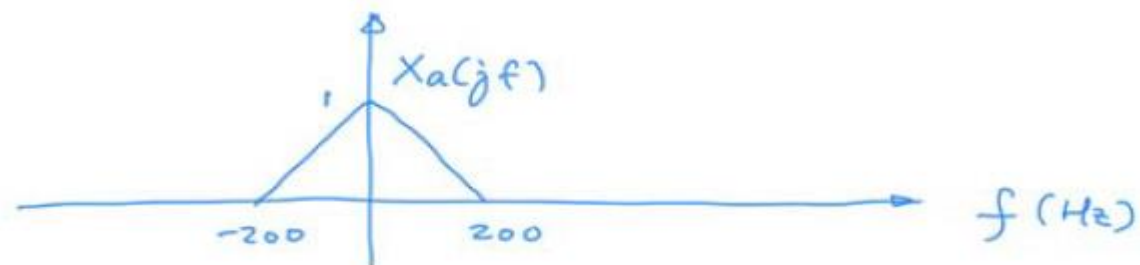
$$X_e(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \left( \sum_{k=-\infty}^{\infty} x[k] \delta[n - 2k] \right) e^{-j\omega n}$$

$$= \sum_{k=-\infty}^{\infty} x[k] \left( \sum_{n=-\infty}^{\infty} \delta[n - 2k] e^{-j\omega n} \right)$$

$$= \sum_{k=-\infty}^{\infty} x[k] e^{-j\omega 2k}$$

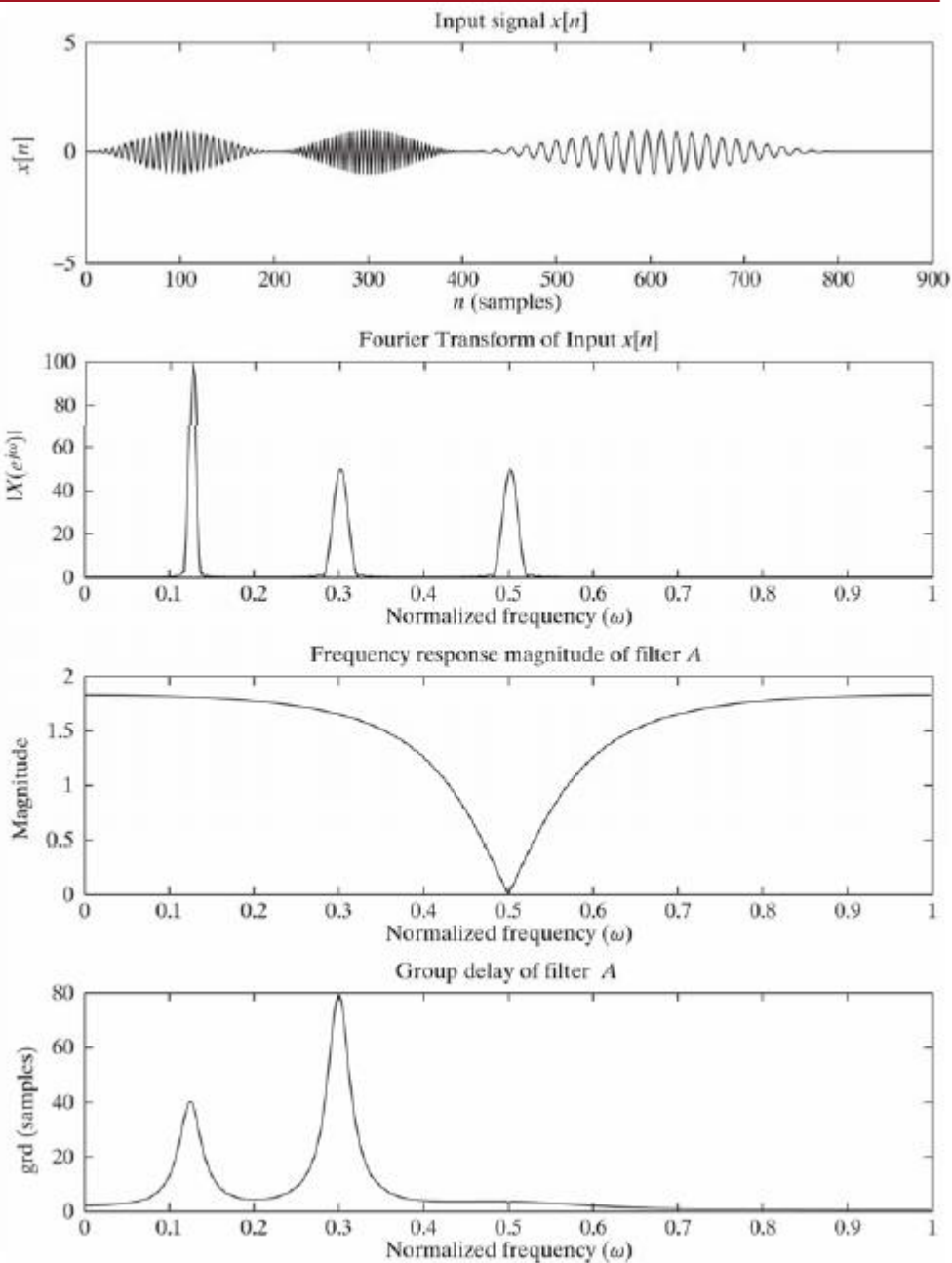
$$= X(e^{j2\omega})$$

Q3.  
(b).



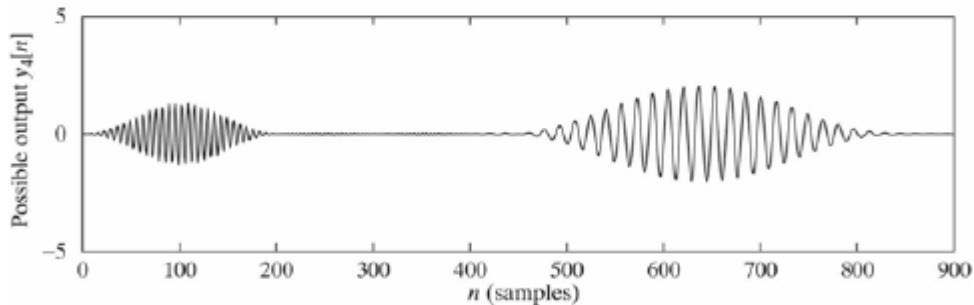
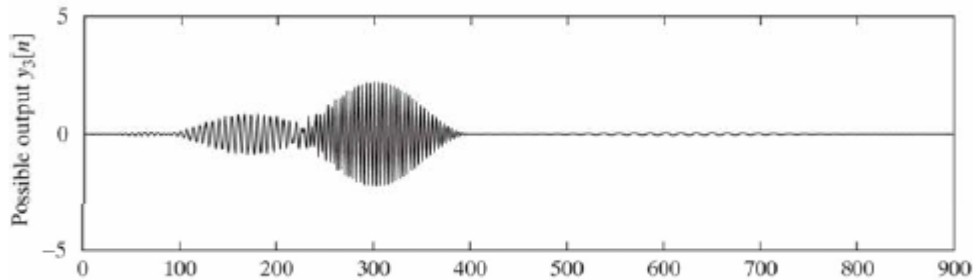
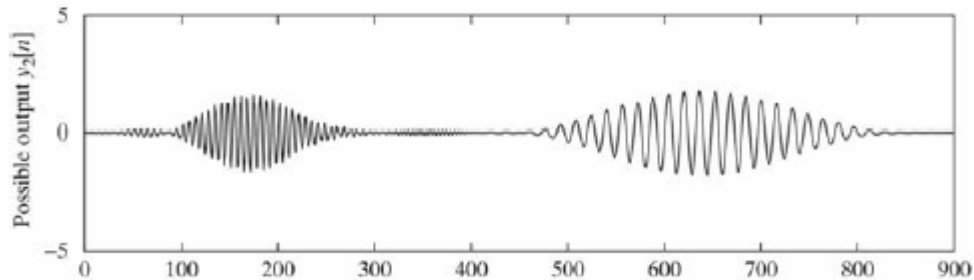
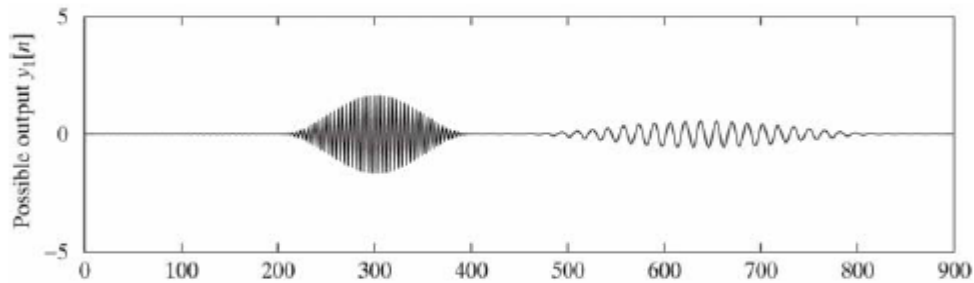


4. [8 marks] A discrete-time LTI system with input  $x[n]$  and output  $y[n]$  has the frequency response magnitude and group delay functions shown in the first figure (which contains  $x[n]$ ,  $|X(e^{j\omega})|$ , Magnitude, and  $\text{grd}$ ). The signal  $x[n]$  is the sum of three narrowband pulses. In particular, the first figure contains the following plots:
- $x[n]$ , the input signal.



- $|X(e^{j\omega})|$ , the Fourier transform magnitude of the particular input  $x[n]$
- Frequency response magnitude plot for the system
- Group delay plot for the system

In the second figure you are given four (4) possible output signals,  $y_i[n]$   $i=1,2,3,4$ . Determine which one of the possible output signals is the output of the system when the input is  $x[n]$ . Provide a justification for your choice.



Q4:

- look at the magnitude response, at  $\omega = 0.5\pi$ , the magnitude = 0. This means that the frequency  $\omega = 0.5\pi$  will be eliminated, so  $y_2[n]$  &  $y_4[n]$  are the candidates.
- group delay for  $\omega = 0.3\pi$  is 80 samples. Group delay for  $\omega = 0.1\pi$  is 40 samples. So  $y_2[n]$  is the answer.



5. [10 marks in total] Consider a causal LTI system satisfying the following difference equation relating  $x[n]$  and  $y[n]$ :

$$y[n] = -\frac{1}{8}(2y[n-1] - 3y[n-2]) + x[n] + 2x[n-1] + x[n-2]$$

- a. Determine the system function  $H(z)$ . Give the zeros and poles. Sketch the pole-zero plot and specify the region of convergence in the z-plane by shading. [8 marks]
- b. Is this system stable? [2 marks]

Q5:

(a)  $y[n] = -\frac{1}{8}(2y[n-1] - 3y[n-2]) + x[n] + 2x[n-1] + x[n-2]$

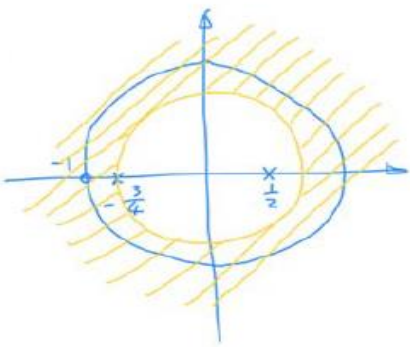
$$Y(z) = -\frac{1}{8}(2Y(z)z^{-1} - 3Y(z)z^{-2}) + X(z) + 2X(z)z^{-1} + X(z)z^{-2}$$

$$Y(z) \left[ 1 + \frac{1}{8}(2z^{-1} - 3z^{-2}) \right] = X(z) [1 + 2z^{-1} + z^{-2}]$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1 + 2z^{-1} + z^{-2}}{1 + \frac{1}{8}(2z^{-1} - 3z^{-2})}$$

$$= \frac{(1+z^{-1})^2 \cdot 8}{8 + 2z^{-1} - 3z^{-2}} = \frac{(1+z^{-1})^2 \cdot 8}{(2-z^{-1})(4+3z^{-1})}$$

$$= \frac{(1+z^{-1})^2}{(1-\frac{1}{2}z^{-1})(1+\frac{3}{4}z^{-1})}$$



(b) stable.