

ELG3175 Introduction to
Communication Systems

Hilbert Transform, Preenvelope and Quadrature Representation of Bandpass Signals



Phase shifting systems and the Hilbert Transform

- Let $x(t)$ be the input to a phase transformer system.
- Let $y(t)$ be the output. Then $y(t)$ is a phase-shifted version of $x(t)$ where all frequency components are shifted by θ .
- Let $x_0(t) = A\cos(2\pi f_0 t)$ be the input.
- Then $y_0(t) = A\cos(2\pi f_0 t + \theta)$.
- Let's change the input frequency. In other words, let $x_1(t) = A\cos(2\pi f_1 t)$.
- Its corresponding output is $y_1(t) = A\cos(2\pi f_1 t + \theta)$.
- The amount of phase shift is independent of input frequency.





Frequency response of phase shifting systems

- For $x_0(t) = A\cos(2\pi f_0 t)$, $X_0(f) = \frac{A}{2} \delta(f - f_0) + \frac{A}{2} \delta(f + f_0)$.
- The output $y_0(t) = A\cos(2\pi f_0 t + \theta)$ has Fourier transform $Y_0(f) = \frac{A}{2} e^{j\theta} \delta(f - f_0) + \frac{A}{2} e^{-j\theta} \delta(f + f_0)$
- Also $X_1(f) = \frac{A}{2} \delta(f - f_1) + \frac{A}{2} \delta(f + f_1)$
- While $Y_1(f) = \frac{A}{2} e^{j\theta} \delta(f - f_1) + \frac{A}{2} e^{-j\theta} \delta(f + f_1)$
- Since the same is true for all input frequencies it is obvious that the frequency response of the phase shifting system is :

$$H(f) = \begin{cases} e^{j\theta}, & f > 0 \\ e^{-j\theta}, & f < 0 \end{cases}$$



Hilbert Transform

- The Hilbert Transform introduces a phase shift of $\theta = -90^\circ$.
- For $x(t)$, its Hilbert transform $x_h(t)$ is a -90° phase shifted version of $x(t)$ ($-\pi/2$ radians).
- The Fourier transform of $x_h(t)$ is $X_h(f)$ which is given by :

$$X_h(f) = \begin{cases} e^{-j\pi/2} X(f), & f > 0 \\ e^{j\pi/2} X(f), & f < 0 \end{cases}$$
$$= -j \operatorname{sgn}(f) X(f)$$



Hilbert Transform

- The Hilbert transform is also given by :

- $$x_h(t) = \mathcal{F}^{-1}\{-j\text{sgn}(f)X(f)\} = x(t) * 1/\pi t$$
$$= \int_{-\infty}^{\infty} \frac{x(\lambda)}{\pi(t-\lambda)} d\lambda = \int_{-\infty}^{\infty} \frac{x(t-\lambda)}{\pi\lambda} d\lambda$$



Examples

- Find the Hilbert Transforms of
 - $x(t) = A\cos(2\pi f_o t)$ et
 - $y(t) = \text{sinc}(t)$

- SOLUTION (a)

$$X(f) = \frac{A}{2} \delta(f - f_o) + \frac{A}{2} \delta(f + f_o) \quad \text{therefore}$$

$$X_h(f) = \frac{-jA}{2} \delta(f - f_o) + \frac{jA}{2} \delta(f + f_o)$$

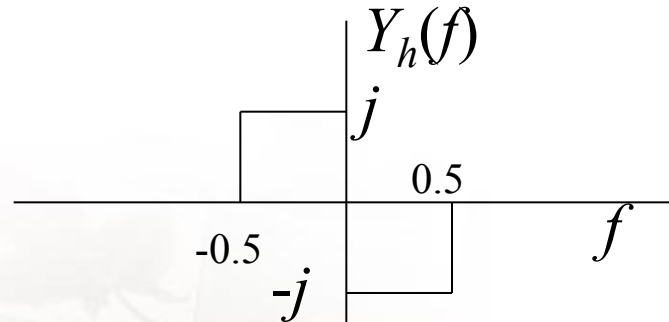
$$\begin{aligned} \text{The Hilbert transform of } x(t) \text{ is } x_h(t) &= \mathcal{F}^{-1}\{X_h(f)\} \\ &= A\sin(2\pi f_o t). \end{aligned}$$





Exemples

- SOLUTION (b)
 - $Y(f) = \Pi(f)$. The Fourier Transform of the Hilbert transform of $y(t)$ is $Y_h(f) = -j\text{sgn}(f)\Pi(f)$.



- $-j\text{sgn}(f)\Pi(f) = -j\Pi(2(f-1/4)) + j\Pi(2(f+1/4))$,

Therefore $y_h(t) = \frac{-j}{2} \text{sinc}(t/2)e^{j\frac{\pi}{2}t} + \frac{j}{2} \text{sinc}(t/2)e^{-j\frac{\pi}{2}t}$

$$= \text{sinc}\left(\frac{t}{2}\right)\sin\left(\frac{\pi}{2}t\right)$$





Positive Pre-envelope

- Let $x(t)$ be a real signal with Fourier Transform $X(f)$.
- Let us define $x_+(t)$ as the positive pre-envelope of $x(t)$.
- The spectrum of $x_+(t)$ is 0 for negative frequencies and proportional to the spectrum of $x(t)$ for positive frequencies.
- The spectrum of the positive pre-envelope is:

$$X_+(f) = \begin{cases} 2X(f), & f > 0 \\ X(0), & f = 0 \\ 0, & f < 0 \end{cases}$$



Positive Pre-envelope

- We can show that $X_+(f) = X(f) + \text{sgn}(f)X(f) = X(f) + j(-j\text{sgn}(f)X(f)) = X(f) + jX_h(f)$ where $X_h(f) = \mathcal{F}\{x_h(t)\}$.
- Therefore

$$x_+(t) = x(t) + jx_h(t)$$





Examples

- Find the positive pre-envelope of $x(t) = \cos(2\pi f_c t)$.
- Find the pre-envelope of $y(t) = \text{sinc}(t)$.
 - SOLUTION
- We know that $x_h(t) = \sin(2\pi f_c t)$, therefore $x_+(t) = \cos(2\pi f_c t) + j\sin(2\pi f_c t)$.
- Therefore $x_+(t) = e^{j2\pi f_c t}$.
- For $y_+(t)$ we should find $Y_+(f)$.
- $Y(f) = \Pi(f)$, therefore $Y_+(f) = 2\Pi(2(f-1/4))$. Therefore $y_+(t) = \mathcal{F}^{-1}\{Y_+(f)\} = \text{sinc}(t/2)e^{j(\pi/2)t}$.





Negative pre-envelope

- The negative pre-envelope of $x(t)$ is a signal that has only the negative spectrum of $x(t)$.

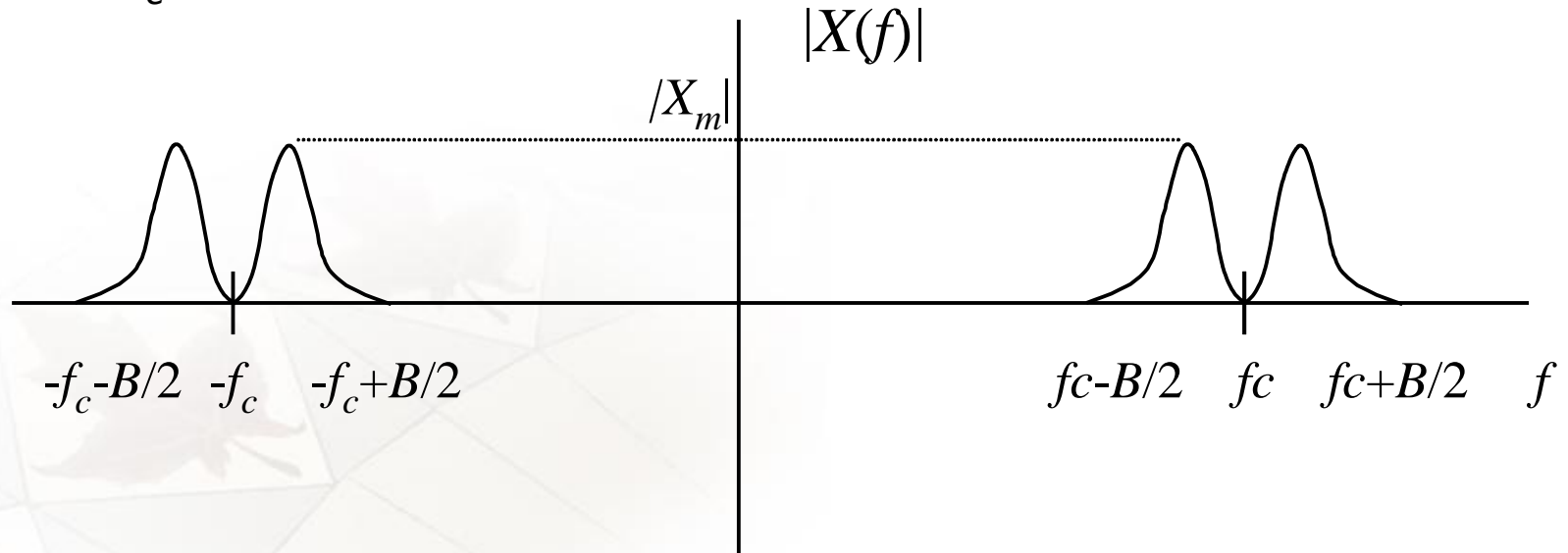
$$X_-(f) = \begin{cases} 2X(f), & f < 0 \\ X(0), & f = 0 \\ 0, & f > 0 \end{cases}$$

- We see that $X_+(f) + X_-(f) = 2X(f)$, therefore $x_+(t) + x_-(t) = 2x(t)$.
- Therefore $x_-(t) = x(t) - jx_h(t)$.



Bandpass Signals

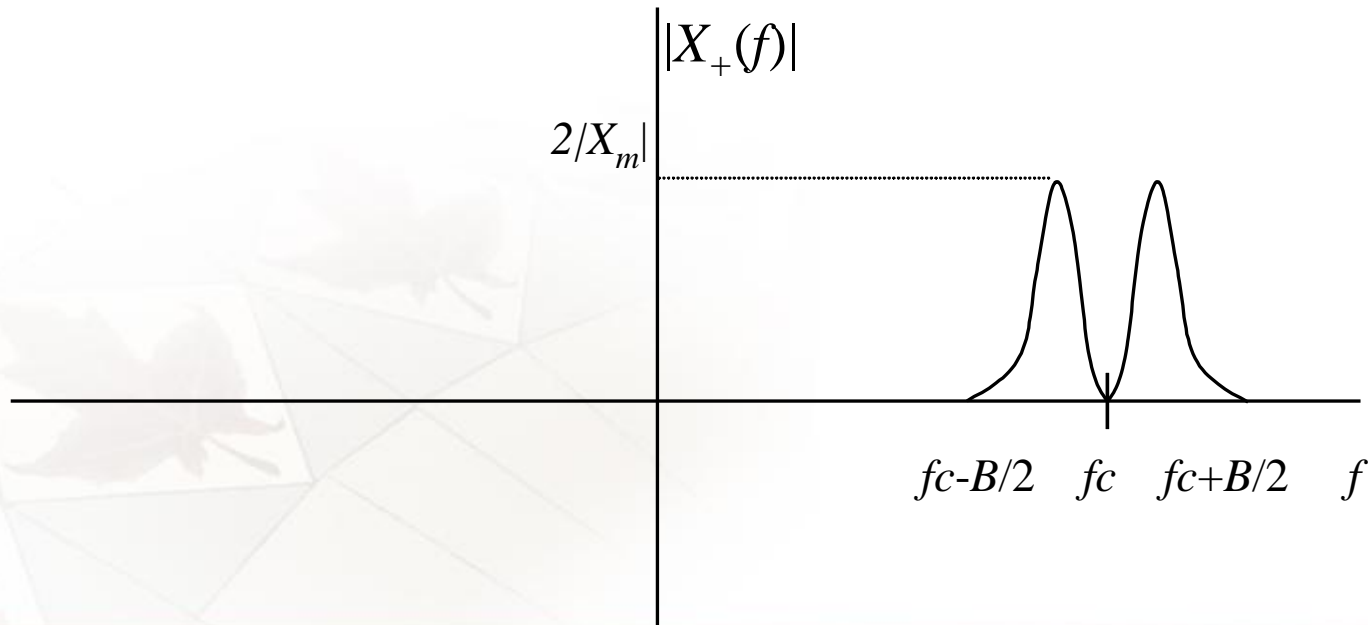
- The signal $x(t)$ is a bandpass signal if its spectrum is non zero in a range of frequencies $f_c - (B/2) \leq |f| \leq f_c + B/2$, and 0 elsewhere where B is its bandwidth and $B < f_c$.





Pre-envelope of a bandpass signal

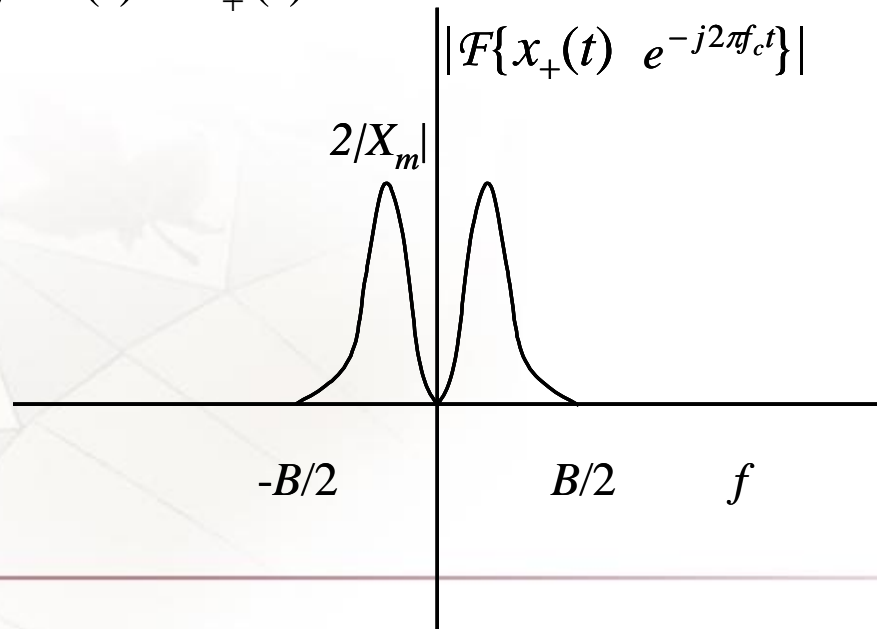
- Let's find the pre-envelope of a bandpass signal, $x(t)$.
- $|X_+(f)|$ is shown below.
- The pre-envelope is $x_+(t) = x(t) + jx_h(t)$.



Complex envelope (lowpass equivalent)



- The complex envelope of $x(t)$, $\tilde{x}(t)$, is its lowpass equivalent
- The spectrum of $\tilde{x}(t)$ has the same form as $x_+(t)$, but it is centred at $f = 0$.
- Therefore, $\tilde{x}(t) = x_+(t)e^{-j2\pi f_c t}$





- Alors, nous définissons $\tilde{x}(t) = x_+(t)e^{-j2\pi f_c t}$ comme l'enveloppe complexe du signal en bande passante $x(t)$.
- Nous voyons du spectre de $\tilde{x}(t)$ que la largeur de bande de l'enveloppe complexe est $B/2$ pour un signal en bande passante avec largeur de bande B .





Quadrature representation of bandpass signals

- If $\tilde{x}(t) = x_+(t)e^{-j2\pi f_c t}$
- Then $x_+(t) = \tilde{x}(t)e^{j2\pi f_c t}$
- Also $x(t) = \text{Re}\{x_+(t)\}$
- Then

$$x(t) = \text{Re}\{\tilde{x}(t)e^{j2\pi f_c t}\} = \text{Re}\{\tilde{x}(t)\} \cos 2\pi f_c t - \text{Im}\{\tilde{x}(t)\} \sin 2\pi f_c t$$

- All bandpass signals can be written in the form

$$x(t) = x_I(t) \cos 2\pi f_c t - x_Q(t) \sin 2\pi f_c t$$

- where $x_I(t) = \text{Re}\{\tilde{x}(t)\}$ and $x_Q(t) = \text{Im}\{\tilde{x}(t)\}$



Example 1

- $x(t) = A\cos(2\pi f_c t + \phi)$.
- Find its complex envelope and its quadrature representation.

– SOLUTION

- $X(f) = (1/2)e^{j\phi}\delta(f-f_c) + (1/2)e^{-j\phi}\delta(f+f_c)$
- $X_+(f) = e^{j\phi}\delta(f-f_c)$

$$\tilde{X}(f) = X_+(f + f_c) = e^{j\phi}\delta(f)$$

$$\tilde{x}(t) = e^{j\phi} = \cos(\phi) + j\sin(\phi)$$

- Therefore $x(t) = x_I(t)\cos(2\pi f_c t) - x_Q(t)\sin(2\pi f_c t) = \cos(\phi)\cos(2\pi f_c t) - \sin(\phi)\sin(2\pi f_c t)$.





Example 2

- $y(t) = 100\sin(2\pi(f_c f_1)t) + 500\cos 2\pi f_c t + 100\sin(2\pi(f_c + f_1)t)$.
- $Y(f) = -j50\delta(f - f_c + f_1) + j50\delta(f + f_c - f_1) + 250\delta(f - f_c) + 250\delta(f + f_c) - j50\delta(f - f_c - f_1) + j50\delta(f + f_c + f_1)$.
- $Y_+(f) = -j100\delta(f - f_c + f_1) + 500\delta(f - f_c) - j100\delta(f - f_c - f_1)$

$$\tilde{Y}(f) = Y_+(f + f_c) = -j100\delta(f + f_1) + 500\delta(f) - j100\delta(f - f_1)$$

$$\tilde{y}(t) = 500 - j200\cos(2\pi f_1 t)$$

- $y(t) = 500\cos(2\pi f_c t) + 200\cos(2\pi f_1 t)\sin(2\pi f_c t)$

