

ELG3175 Introduction to  
Communication Systems  
**Modulation and  
Demodulation of FM  
signals**



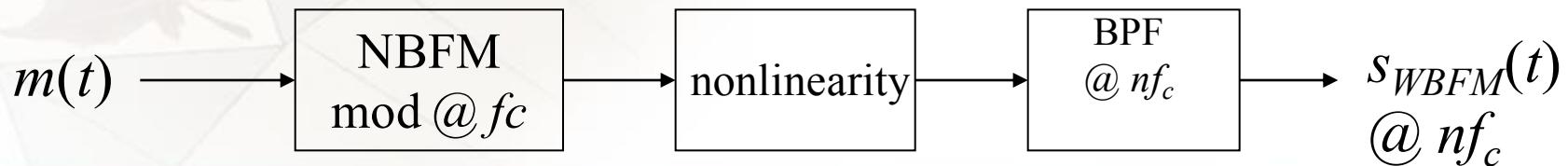
# Generation of WBFM Signals



- Direct method
  - Voltage Controlled Oscillator (VCO)



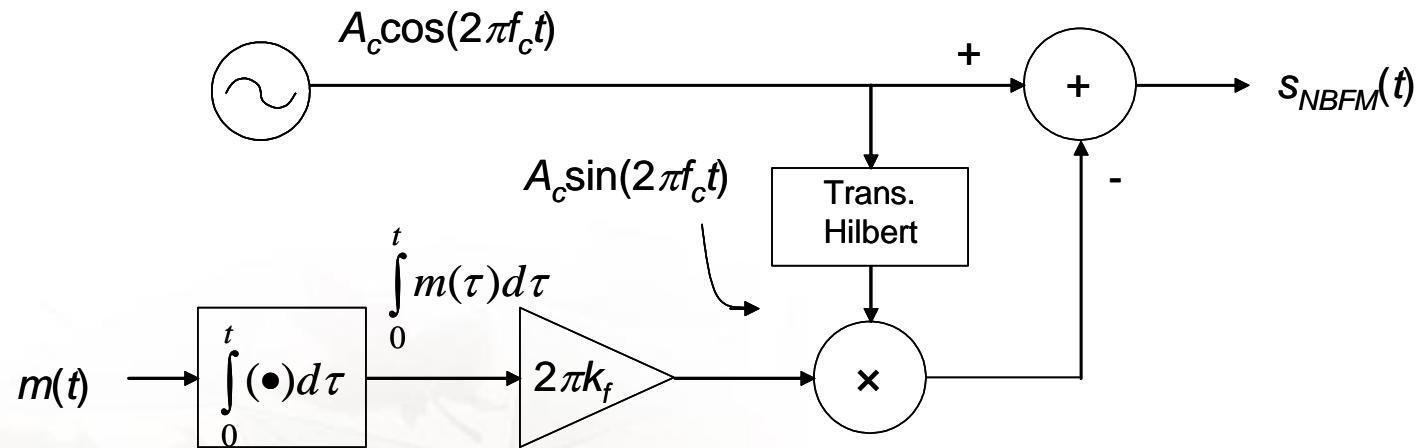
- Indirect method
  - Armstrong's method



# Armstrong's Method



- NBFM Modulator





# Armstrong's method

- Nonlinearity
  - $v_o = a_1 v_i + a_2 v_i^2 + a_3 v_i^3 + \dots$
  - $v_i(t) = s_{NBFM}(t).$
  - Example  $s_{NBFM}(t) = A_c \cos(2\pi f_c t + 2\pi k_f \int m(t) dt) = A_c \cos(\theta_i(t)).$
  - $v_o(t) = a_1 s_{NBFM}(t) + a_2 s_{NBFM}^2(t) + a_3 s_{NBFM}^3(t) \dots$
  - $v_o(t) = a_1 A_c \cos(\theta_i(t)) + a_2 A_c^2 \cos^2(\theta_i(t)) + a_3 A_c^3 \cos^3(\theta_i(t)) \dots$
  - $v_o(t) = a_1 A_c \cos(\theta_i(t)) + a_2 A_c^2 / 2 + (a_2 A_c^2 / 2) \cos(2\theta_i(t)) + (3a_3 A_c^3 / 4) \cos(\theta_i(t)) + (a_3 A_c^3 / 4) \cos(3\theta_i(t)) \dots$
  - $n\theta_i(t) = 2\pi(nf_c)t + 2\pi(nk_f) \int m(t) dt$  (carrier frequency =  $nf_c$  and  $k_f' = nk_f$  therefore  $\beta_F' = n\beta_F$ ).
- BPF is used to pass the spectral component centred @  $f = nf_c$ .



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# Demodulation of FM signals

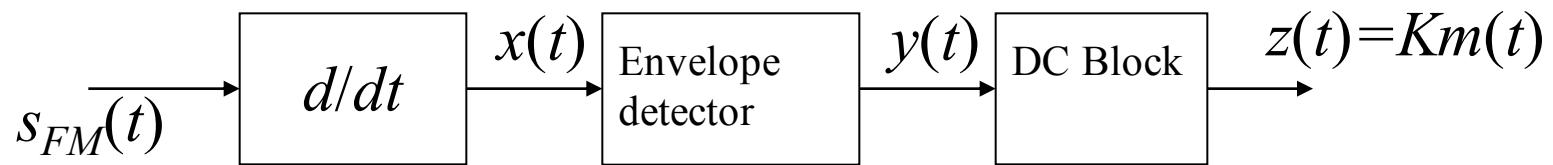


- Differentiator plus envelope detection
- Frequency discriminator.
- Frequency counter.



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# Differentiator and envelope detector



# Differentiator and envelope detector



$$\begin{aligned}x(t) &= \frac{ds_{FM}(t)}{dt} \\&= \frac{d}{dt}(A_c \cos(\theta_i(t))) \\&= -\frac{d\theta_i(t)}{dt} A_c \sin(\theta_i(t)) \\&= 2\pi A_c f_i(t) \sin\left(2\pi f_c t + 2\pi k_f \int m(t) dt + \pi\right) \\&= 2\pi A_c (f_c + k_f m(t)) \sin\left(2\pi f_c t + 2\pi k_f \int m(t) dt + \pi\right)\end{aligned}$$

$f_c \gg |k_f m(t)|$  alors  $2\pi A_c (f_c + k_f m(t)) > 0$ .



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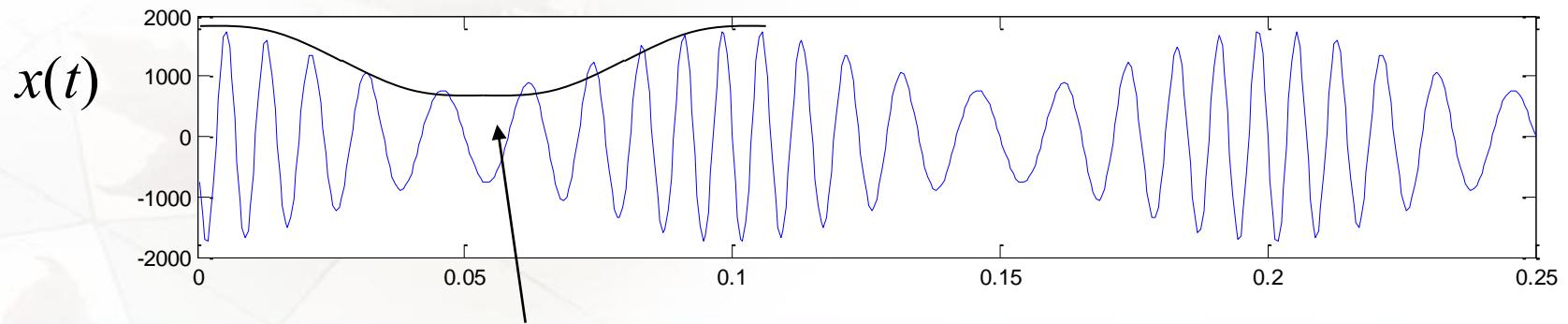
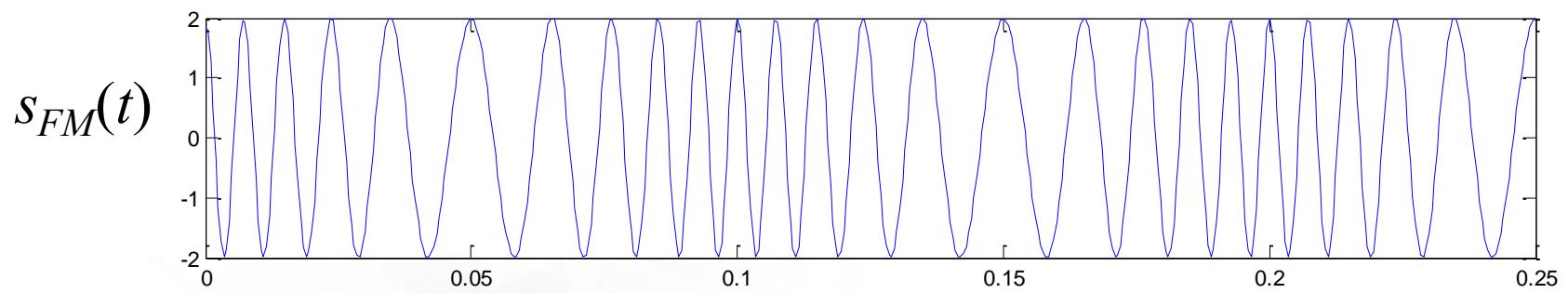
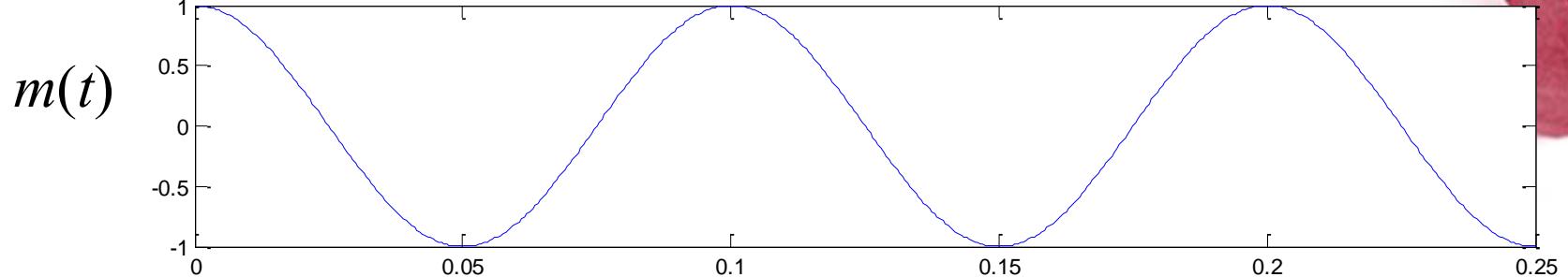
# Example



- $m(t) = \cos 2\pi 10t, f_c = 100, A_c = 2, k_f = 40 \text{ Hz/V}.$
- $s_{FM}(t) = 2\cos(2\pi 200t + 4\sin 2\pi 10t)$
- $x(t) = 4\pi(100 + 40\cos 2\pi 10t)\sin(2\pi 100t + 4\sin 2\pi 10t + \pi)$



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$$2\pi A_c(f_c + k_f m(t))$$



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# Differentiator and envelope detector



- Output of envelope detector
  - $y(t) = 2\pi A_c(f_c + k_f m(t)) = 2\pi A_c f_c + 2\pi A_c k_f m(t)$
  - Assuming that  $m(t)$  has no DC component ( $M(f) = 0$  for  $f = 0$ ), then
- Output of DC block
  - $z(t) = 2\pi A_c k_f m(t) = Km(t).$

# Fluctuations in received amplitude



- The power of the received signal is  $A^2/2$ .
- The received power is inversely proportional to the square of the distance between the transmitter and the receiver.
- The power can also fluctuate due to conditions between the transmitter and receiver (rain, obstructions etc)
- This variation in received power causes fluctuations in the received amplitude.
- $r(t) = A(t)\cos(\theta_i(t))$ .



# Differentiator and envelope detector when received amplitude fluctuates



$$\begin{aligned}x(t) &= \frac{dr(t)}{dt} \\&= \frac{d}{dt}(A(t)\cos(\theta_i(t))) \\&= -\frac{d\theta_i(t)}{dt} A(t)\sin(\theta_i(t)) + \frac{dA(t)}{dt}\cos(\theta_i(t)) \\&= 2\pi A(t) f_i(t) \sin\left(2\pi f_c t + 2\pi k_f \int m(t) dt + \pi\right) \\&= 2\pi A(t) \left(f_c + k_f m(t)\right) \sin\left(2\pi f_c t + 2\pi k_f \int m(t) dt + \pi\right) \\&\quad + \frac{dA(t)}{dt} \cos\left(2\pi f_c t + 2\pi k_f \int m(t) dt\right)\end{aligned}$$



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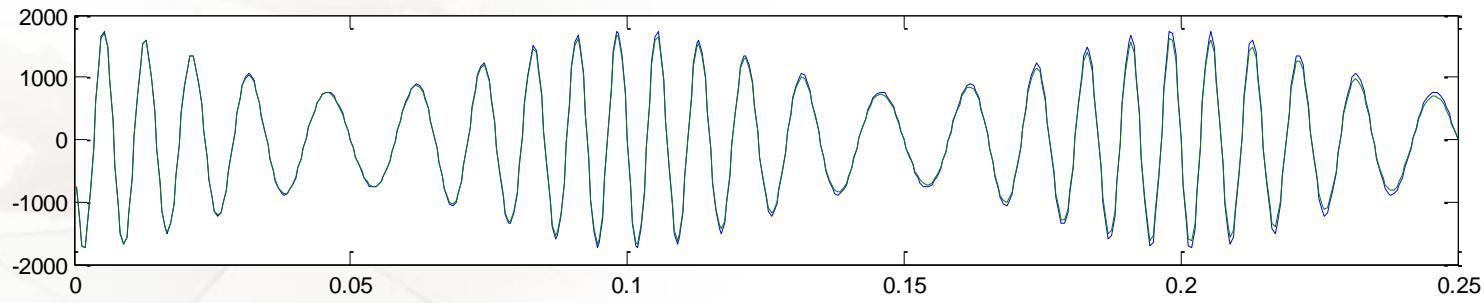
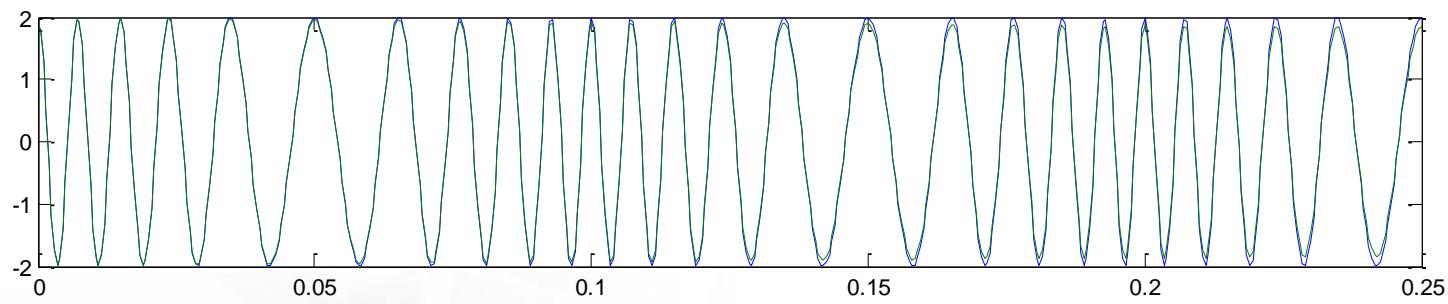
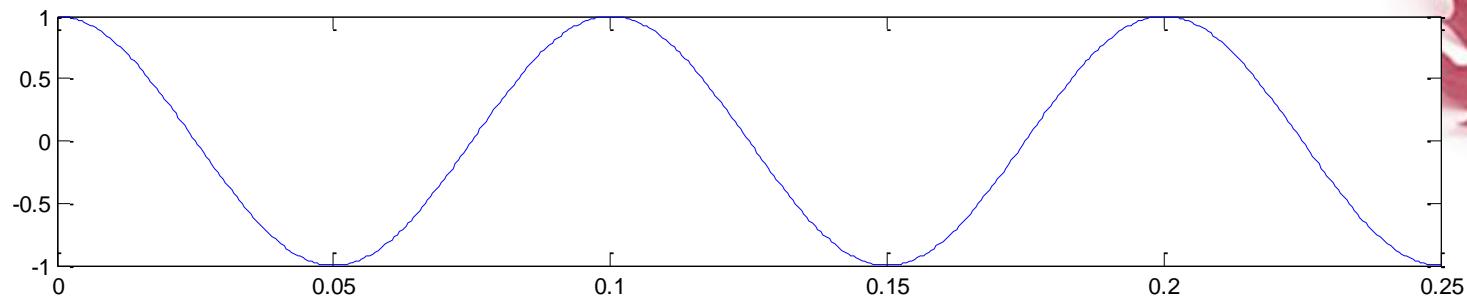
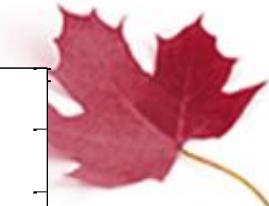
# Examples



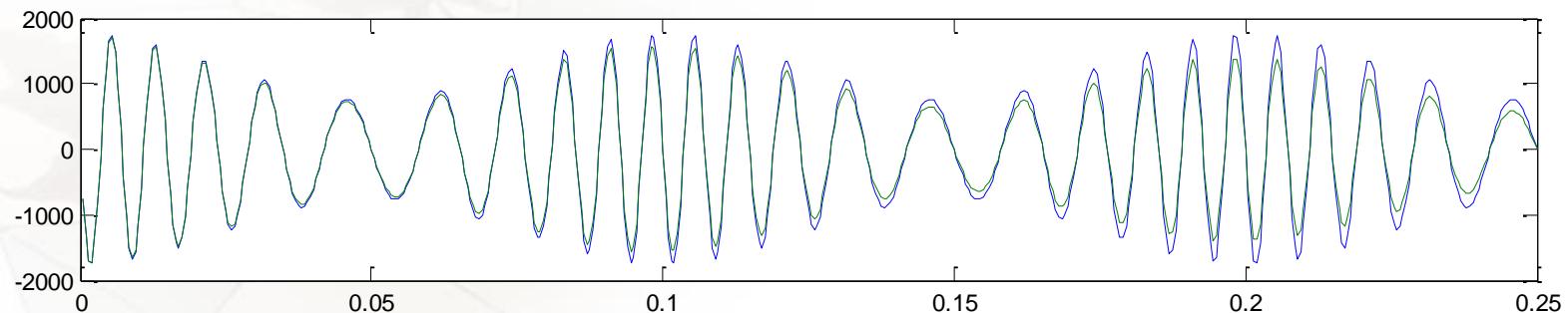
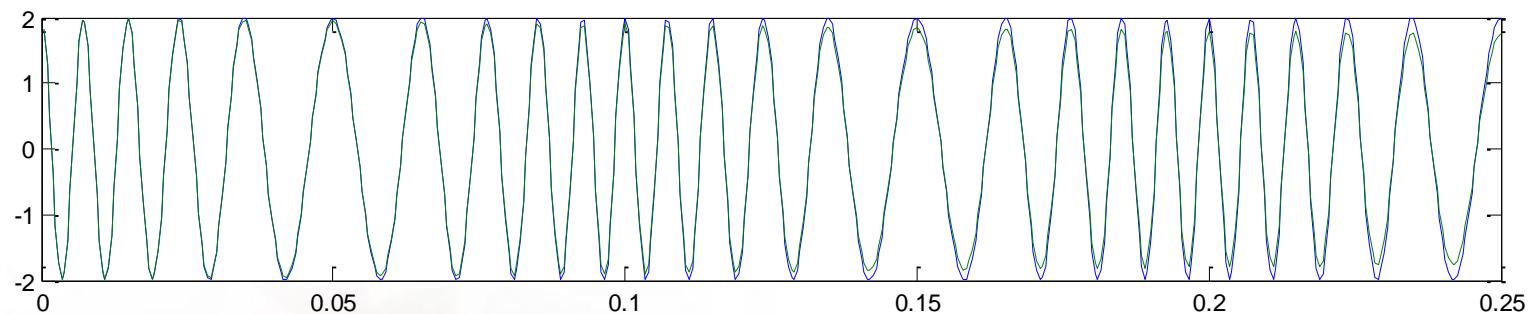
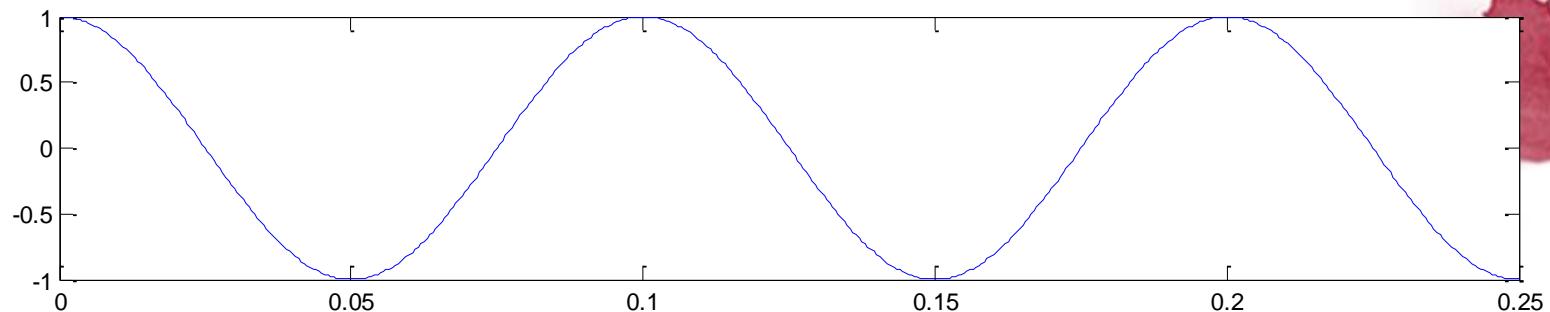
- Example 1
- $m(t) = \cos 2\pi 10t, f_c = 100, A(t) = 2e^{-t/3}, k_f = 40 \text{ Hz/V}.$
- $s_{FM}(t) = 2\cos(2\pi 200t + 4\sin 2\pi 10t)$
- $x(t) = 4\pi e^{-t/3}(100 + 40\cos 2\pi 10t)\sin(2\pi 100t + 4\sin 2\pi 10t + \pi) - (2/3)e^{-t/3}\cos(2\pi 200t + 4\sin 2\pi 10t)$
- Example 2
- $m(t) = \cos 2\pi 10t, f_c = 100, A(t) = 2(1-t), k_f = 40 \text{ Hz/V}.$
- $s_{FM}(t) = 2\cos(2\pi 200t + 4\sin 2\pi 10t)$
- $x(t) = 4\pi(1-t)(100 + 40\cos 2\pi 10t)\sin(2\pi 100t + 4\sin 2\pi 10t + \pi) - 2t\cos(2\pi 200t + 4\sin 2\pi 10t)$



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# Conclusion



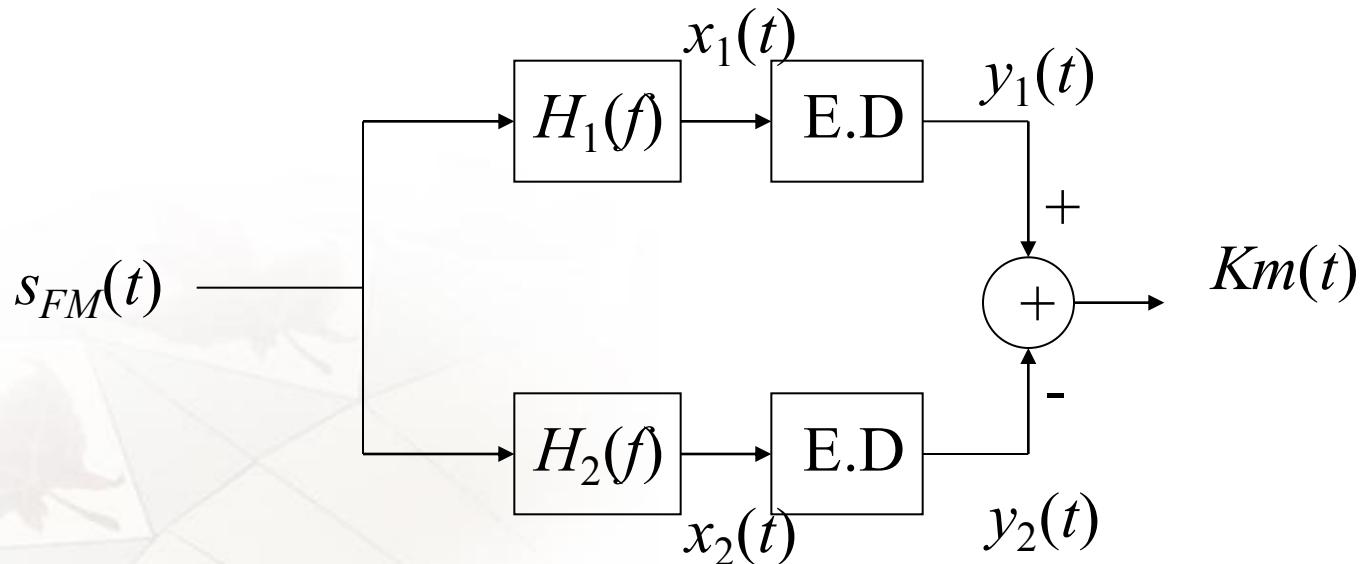
- When the amplitude varies the output of the envelope detector will be a distorted version of  $m(t)$ .
- Solution: 1) passband limiter to make the received amplitude a constant , 2) detector that does not require taking the derivative (frequency counter of Lab 3).
- Another problem with the differentiator is that the output has a very high amplitude due to the magnitude of  $f_c$ .
  - Frequency discriminator



# Frequency discriminator



- Similar to differentiator
- Input to envelope detector has lower amplitude.

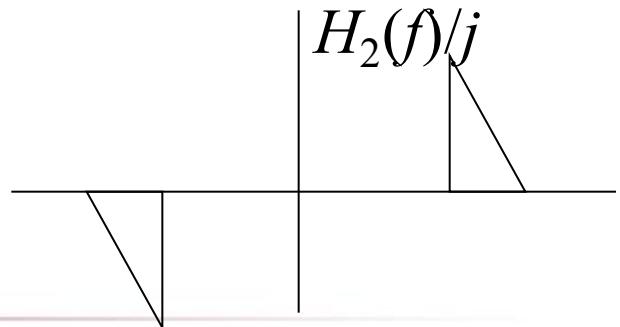
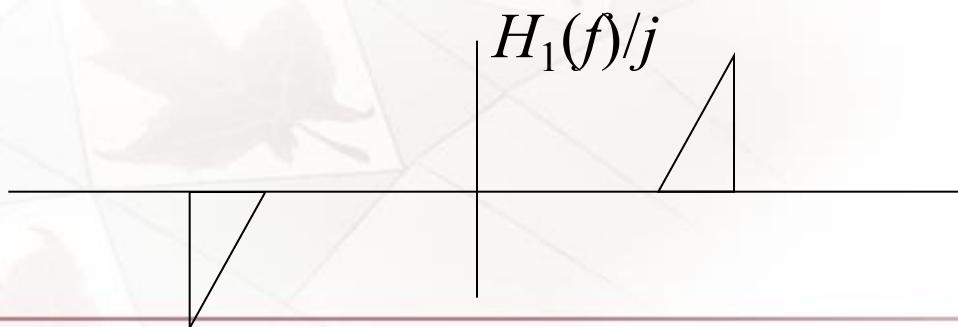


# $H_1(f)$ and $H_2(f)$



$$H_1(f) = \begin{cases} j2\pi a(f - f_c + \frac{B}{2}), & f_c - \frac{B}{2} < f < f_c + \frac{B}{2} \\ j2\pi a(f + f_c - \frac{B}{2}), & -f_c - \frac{B}{2} < f < -f_c + \frac{B}{2} \\ 0, & \text{otherwise} \end{cases}$$

$$H_2(f) = \begin{cases} -j2\pi a(f - f_c - \frac{B}{2}), & f_c - \frac{B}{2} < f < f_c + \frac{B}{2} \\ -j2\pi a(f + f_c + \frac{B}{2}), & -f_c - \frac{B}{2} < f < -f_c + \frac{B}{2} \\ 0, & \text{otherwise} \end{cases}$$



# $S_{FM}(f)H_1(f)$ and its complex envelope



- $X_1(f) = S_{FM}(f)H_1(f) = (1/2)S_+(f)j2\pi a(f-f_c+B/2) + (1/2)S_-(f)j2\pi a(f+f_c-B/2)$
- The FT of the complex envelope of  $x_1(t)$  is:

$$\begin{aligned}\tilde{X}_1(f) &= S_+(f + f_c)j2\pi a(f + \frac{B}{2}) \\ &= \tilde{S}_{FM}(f)j2\pi a(f + \frac{B}{2}) \\ &= aj2\pi f\tilde{S}_{FM}(f) + aj2\pi \frac{B}{2}\tilde{S}_{FM}(f) \\ \tilde{x}_1(t) &= a\left(\frac{d\tilde{S}_{FM}(t)}{dt} + j\pi B\tilde{S}_{FM}(t)\right)\end{aligned}$$

similarly

$$\tilde{x}_2(t) = a\left(\frac{d\tilde{S}_{FM}(t)}{dt} + j\pi B\tilde{S}_{FM}(t)\right)$$



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$$s_+(t) = A_c e^{j\theta_i(t)} = A_c e^{j(2\pi f_c t + 2\pi k_f \int m(t) dt)}$$



$$\tilde{s}_{FM}(t) = A_c e^{j2\pi k_f \int m(t) dt}$$

$$\frac{d\tilde{s}_{FM}(t)}{dt} = j2\pi k_f A_c m(t) e^{j2\pi k_f \int m(t) dt}$$

$$\tilde{x}_1(t) = aj\pi A_c B \left[ 1 + \frac{2k_f m(t)}{B} \right] e^{j2\pi k_f \int m(t) dt}$$

$$\tilde{x}_2(t) = aj\pi A_c B \left[ 1 - \frac{2k_f m(t)}{B} \right] e^{j2\pi k_f \int m(t) dt}$$

$$x_1(t) = \pi A_c B a \left[ 1 + \frac{2k_f m(t)}{B} \right] \cos \left( 2\pi f_c t + 2\pi k_f \int m(t) dt + \frac{\pi}{2} \right)$$

$$x_2(t) = \pi A_c B a \left[ 1 - \frac{2k_f m(t)}{B} \right] \cos \left( 2\pi f_c t + 2\pi k_f \int m(t) dt + \frac{\pi}{2} \right)$$

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$$y_1(t) - y_2(t) = 4\pi A_c a k_f m(t) = Km(t)$$



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